

SIGNAL SUBSPACES, ORTHOGONAL PROJECTIONS, AND LEAST SQUARES ESTIMATION

The Signal Subspace Model

Let $\underline{a}_1, \dots, \underline{a}_p \in \mathbb{R}^N$ (or \mathbb{C}^N) be linearly independent (so $p \leq n$), and consider the $N \times p$ matrix

$$A = [\underline{a}_1 \ \dots \ \underline{a}_p].$$

Let $\langle A \rangle$ denote the linear span of the columns of A (equivalently, the image of A). Then

$$\dim(\langle A \rangle) = \text{rank}(A) = p$$

$$\dim(\langle A \rangle^\perp) = n-p$$

Let $\underline{b}_1, \dots, \underline{b}_{N-p} \in \mathbb{R}^N$ (or \mathbb{C}^N) be a basis for $\langle A \rangle^\perp$ and set

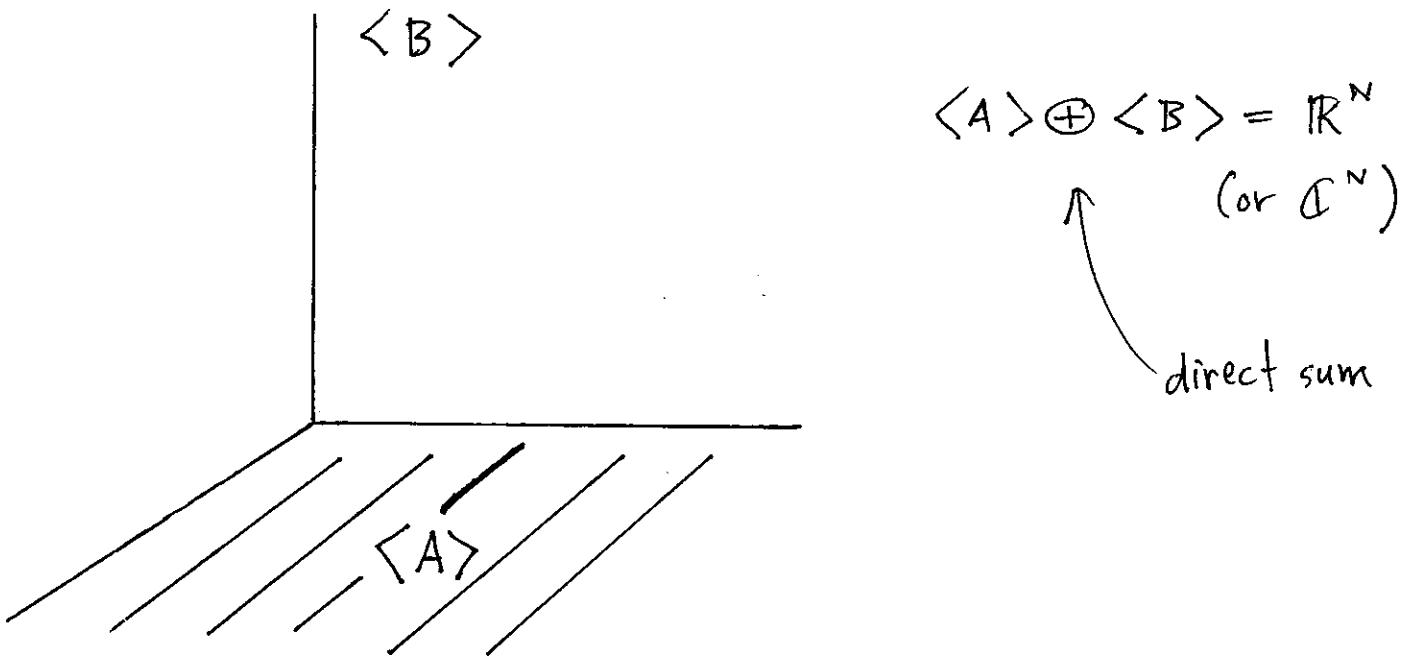
$$B = [\underline{b}_1 \ \dots \ \underline{b}_{N-p}] \quad N \times (N-p)$$

Then

$$\langle \underline{a}_i, \underline{b}_j \rangle = 0 \quad \begin{matrix} i=1, \dots, p \\ j=1, \dots, N-p \end{matrix}$$

and $\{\underline{a}_1, \dots, \underline{a}_p, \underline{b}_1, \dots, \underline{b}_{N-p}\}$ is a basis for \mathbb{R}^N (\mathbb{C}^N)

The subspaces $\langle A \rangle$ and $\langle B \rangle$ form an orthogonal decomposition of \mathbb{R}^N (or \mathbb{C}^N)



Note] The vectors $\underline{q}_1, \dots, \underline{q}_p$ are not necessarily orthogonal among themselves. The same goes for $\underline{b}_1, \dots, \underline{b}_{N-p}$.

(a) In the , we assume our observed signal \underline{x} has the form

$$\underline{x} = \underline{\Sigma} + \underline{w}$$

where

$\underline{\Sigma} \in \langle A \rangle$ is the signal of interest

\underline{w} is entirely noise.

We use the following terminology:

$$\langle A \rangle =$$

(b)

$$\langle B \rangle =$$

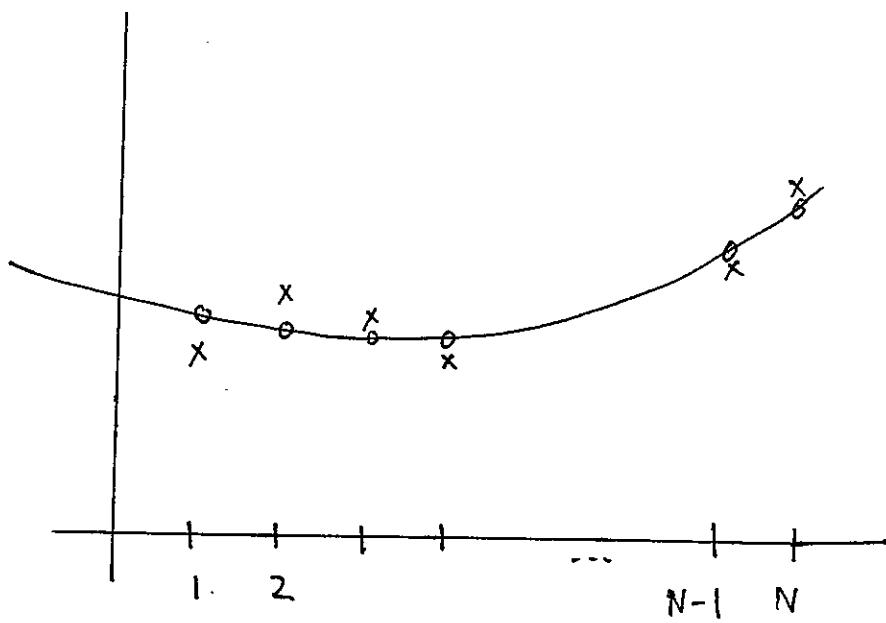
even though $w \notin \langle B \rangle$ in general.

Example] Quadratic polynomial model

$$s(n) = \theta_2 n^2 + \theta_1 n + \theta_0, \quad n = 1, \dots, N$$

$\Rightarrow \underline{s} = A \underline{\theta}$ where

$$A = \begin{bmatrix} 1 & 1 & 1 \\ 1 & 2 & 4 \\ \vdots & \vdots & \vdots \\ 1 & N & N^2 \end{bmatrix} \quad \underline{\theta} = \begin{bmatrix} \theta_0 \\ \theta_1 \\ \theta_2 \end{bmatrix} \in \mathbb{R}^3$$



○ → clean signal (unobserved)

x → noisy signal (observed)

Exercise Consider the sinusoidal signal

$$s(n) = D \cdot \cos(2\pi f n + \phi), \quad n=0, \dots, N-1$$

where f is known but D, ϕ are unknown.

Express $\underline{s} = [s(0) \ \dots \ s(N-1)]^T$ as an element in a two-dimensional subspace.

That is, write

$$\underline{s} = A \cdot \underline{\theta}$$

where A is a known $N \times 2$ matrix and $\underline{\theta}$ is unknown.

Solution 1] Use $\cos(\alpha) = \frac{e^{j\alpha} + e^{-j\alpha}}{2}$. Then

$$s(n) = \underbrace{\left(\frac{D}{2}e^{j\phi}\right)}_{\theta_1} e^{2\pi jfn} + \underbrace{\left(\frac{D}{2}e^{-j\phi}\right)}_{\theta_2} e^{-2\pi jfn}$$

$$\Rightarrow \underline{\Sigma} = \begin{bmatrix} 1 & 1 \\ e^{2\pi j f} & e^{-2\pi j f} \\ e^{4\pi j f} & e^{-4\pi j f} \\ \vdots & \vdots \\ e^{2(N-1)\pi j f} & e^{-2(N-1)\pi j f} \end{bmatrix} \begin{bmatrix} \theta_1 \\ \theta_2 \end{bmatrix}$$

$$\underline{q}_1, \underline{q}_2 \in \mathbb{C}^N, \quad \underline{\theta} \in \mathbb{C}^2$$

Solution 2] Use $\cos(\alpha + \beta) = \cos(\alpha)\cos(\beta) - \sin(\alpha)\sin(\beta)$

$$s(n) = \underbrace{\left(D \cos(\phi)\right)}_{\theta_1} \cos(2\pi fn) + \underbrace{\left(-D \sin(\phi)\right)}_{\theta_2} \sin(2\pi fn)$$

$$\Rightarrow \underline{\Sigma} = \begin{bmatrix} \cos(1\pi f) & 0 \\ \cos(2\pi f) & \sin(2\pi f) \\ \cos(4\pi f) & \sin(4\pi f) \\ \vdots & \vdots \\ \cos(2(N-1)\pi f) & \sin(2(N-1)\pi f) \end{bmatrix} \begin{bmatrix} \theta_1 \\ \theta_2 \end{bmatrix}$$

$$\underline{q}_1, \underline{q}_2 \in \mathbb{R}^N, \quad \underline{\theta} \in \mathbb{R}^2 \quad \text{if } D \in \mathbb{R}$$

Orthogonal Projection

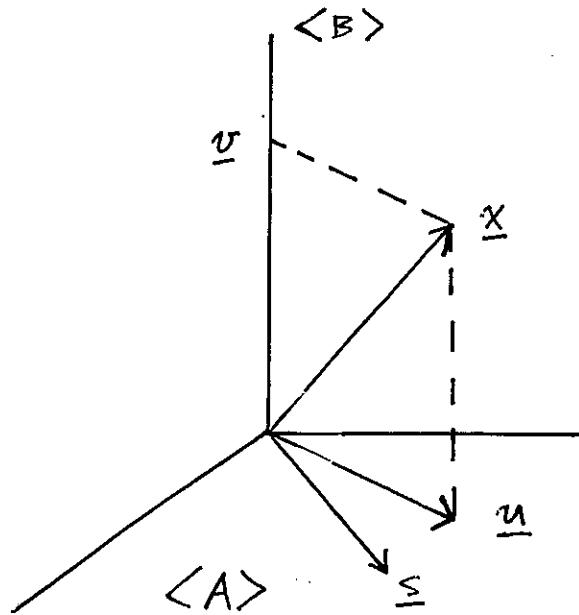
How can we use the knowledge of $\langle A \rangle$ to estimate $\underline{\Sigma}$ from $\underline{x} = \underline{\Sigma} + \underline{w}$?

Since $\langle A \rangle$ and $\langle B \rangle$ are orthogonal complements, we can uniquely write

$$\underline{x} = \underline{u} + \underline{v}$$

where $\underline{u} \in \langle A \rangle$ and $\underline{v} \in \langle B \rangle$.

Since \underline{v} is pure noise, it makes sense to remove it.



In general, $\underline{\Sigma} \neq \underline{u}$ because \underline{w} has some component in $\langle A \rangle$.

The operator that maps $\underline{x} \mapsto \underline{u}$ is called
 ③ the A^H onto $\langle A \rangle$ and is denoted Π_A .

Proposition

$$\Pi_A = A(A^H A)^{-1} A^H$$

Proof] Let $\underline{x} = \underline{u} + \underline{v}$, $\underline{u} \in \langle A \rangle$, $\underline{v} \in \langle B \rangle$.

We must show $A(A^H A)^{-1} A^H \underline{x} = \underline{u}$. Since the columns of A and B form a basis, we can write

$$\underline{u} = A \underline{\theta}, \quad \underline{\theta} \in \mathbb{R}^{p \times 1}$$

$$\underline{v} = B \underline{\phi}, \quad \underline{\phi} \in \mathbb{R}^{(N-p) \times 1}$$

Then

$$\begin{aligned}
 & A(A^H A)^{-1} A^H \underline{x} \\
 &= A \underbrace{(A^H A)^{-1} A^H A \underline{\theta}}_{I_{p \times p}} + A \underbrace{(A^H A)^{-1} A^H B \underline{\phi}}_{O_{p \times (N-p)}} \\
 &= A \underline{\theta} \\
 &= \underline{u} \quad \blacksquare
 \end{aligned}$$

Properties of projections

- $\Pi_A^H =$ "self-adjoint"
- $\Pi_A^2 =$ "idempotent"
- (d) • $\Pi_A + \Pi_B =$
- $\Pi_A \cdot \Pi_B =$
- If q_1, \dots, q_p are orthonormal, then

$$\Pi_A = AA^H$$

Filtering interpretation

The projection operator is analogous to a bandpass filter; we only retain that information which resides in the passband, which corresponds to the signal subspace.

Least Squares Estimation

To estimate $\underline{s} = A\underline{\theta}$ where

$$\underline{x} = \underline{s} + \underline{w}$$

we use the projection onto $\langle A \rangle$:

$$\begin{aligned}\hat{\underline{s}} &= \text{Pr}_A \underline{x} \\ &= A(A^H A)^{-1} A^H \underline{x}.\end{aligned}$$

What if we want to estimate $\underline{\theta}$?

An estimate $\hat{\underline{\theta}}$ of $\underline{\theta}$ should satisfy

$$\hat{\underline{s}} = A\hat{\underline{\theta}}.$$

Therefore, an obvious estimate is

②

$$\hat{\underline{\theta}} =$$

It turns out that this is the solution to the least squares problem.

Proposition] The unique solution of

$$\min_{\underline{\theta}} \|\underline{x} - A\underline{\theta}\|^2 \quad (\underline{\theta} \in \mathbb{R}^P \text{ or } \mathbb{C}^P)$$

is $\hat{\underline{\theta}} = (A^H A)^{-1} A^H \underline{x}$.

Proof] Write $\underline{x} = \underline{u} + \underline{v}$ where $\underline{u} \in \langle A \rangle$ and $\underline{v} \in \langle A \rangle^\perp$. Observe

$$\begin{aligned} \|\underline{x} - A\underline{\theta}\|^2 &= \|\underline{u} - A\underline{\theta} + \underline{v}\|^2 \\ &= \langle \underline{u} - A\underline{\theta} + \underline{v}, \underline{u} - A\underline{\theta} + \underline{v} \rangle \\ &= \langle \underline{u} - A\underline{\theta}, \underline{u} - A\underline{\theta} \rangle + \langle \underline{v}, \underline{v} \rangle \\ &\quad + \underbrace{\langle \underline{u} - A\underline{\theta}, \underline{v} \rangle}_{=0} + \underbrace{\langle \underline{v}, \underline{u} - A\underline{\theta} \rangle}_{=0} \\ &= \|\underline{u} - A\underline{\theta}\|^2 + \|\underline{v}\|^2. \end{aligned}$$

The second term is independent of $\underline{\theta}$. Therefore, to minimize the expression, the best we can do is to make the first term 0 by taking

$$\underline{\theta} = \hat{\underline{\theta}} = (A^H A)^{-1} A^H \underline{x}.$$

Then $A\hat{\underline{\theta}} = \pi_A \underline{x} = \underline{u}$. To see that $\hat{\underline{\theta}}$ is unique, if $\underline{\theta}'$ is also such that $\|\underline{u} - A\underline{\theta}'\| = 0$, then

$$A\underline{\theta}' = \underline{u} \Rightarrow A\underline{\theta}' = A\hat{\underline{\theta}}$$

$$\Rightarrow \underline{\theta}' = \hat{\underline{\theta}}$$

since the columns of A are linearly independent.

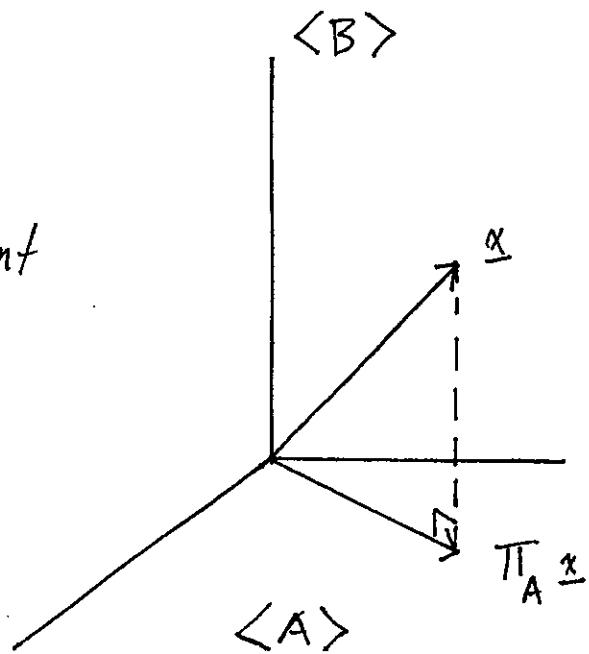
□

Minimum distance property

We may conclude that

$\pi_A \underline{x}$ is the unique point

in $\langle A \rangle$ that is closest
to \underline{x} .



Remark] The operator

$$A^\# := (A^H A)^{-1} A^H$$

is called the pseudo-inverse of A .

Example Recall the sinusoid

$$s(n) = D \cdot \cos(2\pi f n + \phi), \quad n = 0, \dots, N-1$$

where f is known and D, ϕ are unknown.

How can we estimate D and ϕ from \underline{x} ?

Complex solution: $\underline{s} = A \underline{\theta}$ where

$$\underline{\theta} = \begin{bmatrix} \frac{D}{2} e^{j\phi} \\ \frac{D}{2} e^{-j\phi} \end{bmatrix}. \quad \text{Use pseudo-inverse to}$$

compute $\hat{\underline{\theta}}$ and form

$$\hat{D} = \sqrt{4 \cdot \hat{\theta}_1 \cdot \hat{\theta}_2}, \quad \hat{\phi} = \frac{1}{2} \arg\left(\frac{\hat{\theta}_1}{\hat{\theta}_2}\right)$$

Real solution: $\underline{s} = A \underline{\theta}$ where

$$\underline{\theta} = \begin{bmatrix} D \cos(\phi) \\ -D \sin(\phi) \end{bmatrix}. \quad \text{Use pseudo-inverse to}$$

compute $\hat{\underline{\theta}}$ and form

$$\hat{D} = \sqrt{\hat{\theta}_1^2 + \hat{\theta}_2^2}, \quad \hat{\phi} = \tan^{-1}\left(-\frac{\hat{\theta}_2}{\hat{\theta}_1}\right)$$

(assuming D to be real)

Summary

- If a signal lies in a subspace, it can be estimated by projection onto that subspace.
- This "filters out" any noise in the noise subspace.
- The projection satisfies the minimum distance property, and is closely related to the least squares problem.
- This approach is non-statistical because no probability model is specified for the noise. Yet it turns out to be equivalent or similar to many methods we will see later.

Key

- a. signal subspace model
- b. signal subspace, noise subspace
- c. orthogonal projection
- d. Π_A , Π_A^\perp , $I_{N \times N}$, $O_{N \times N}$
- e. $(A^H A)^{-1} A^H \mathbf{x}$