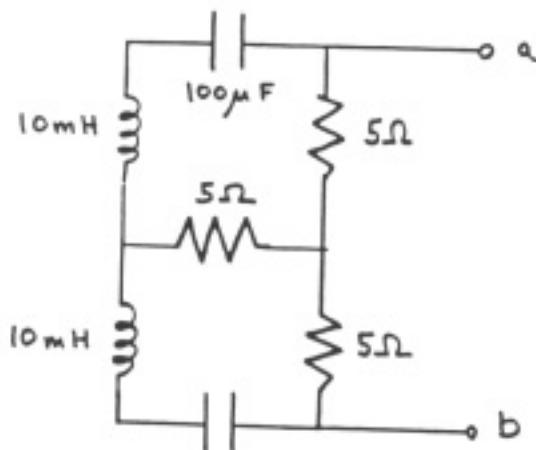


F99 EEC S 210  
Midterm #2 Solutions

Problem: 1 (20 pts)

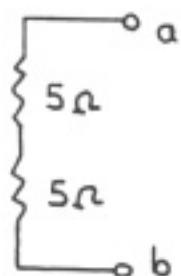
- 1.a. (5 pts) Consider the circuit indicated below. Find the impedance across terminals  $a, b$  of the circuit for  $\omega = 0$  (D.C.) and the impedance for  $\omega = \infty$ .



$$Z_C = \frac{1}{j\omega C} = -j\infty \text{ @ } \omega = 0$$

$$Z_L = j\omega L = j\infty \text{ @ } \omega = \infty$$

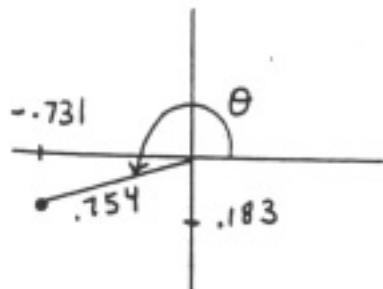
$\therefore$  at either  $\omega = 0$  or  $\omega = \infty$  series combination of  $L \& C$  has  $|Z_{L+C}| = \infty \Rightarrow$  open-circuit. Thus the above circuit reduces to



$$Z_{\omega=0} = \underline{\hspace{2cm} 10 \Omega \hspace{2cm}}, \quad Z_{\omega=\infty} = \underline{\hspace{2cm} 10 \Omega \hspace{2cm}}$$

1.b. (5 pts) Express the complex variable  $W$  below in both polar and rectangular form

$$\begin{aligned}
 W &= \frac{e^{j\pi/3} - 2e^{j\pi/4}}{1+j} \\
 &= \frac{e^{j\pi/3} - 2e^{j\pi/4}}{\sqrt{2} e^{j\pi/4}} \\
 &= \frac{1}{\sqrt{2}} e^{j\pi/12} - \sqrt{2} \\
 &= \frac{1}{\sqrt{2}} (\cos \pi/12 + j \sin \pi/12) - \sqrt{2} \\
 &= .683 + j.183 - \sqrt{2} \\
 &= -.731 + j.183
 \end{aligned}$$



$$\begin{aligned}
 r &= [(0.731)^2 + (.183)^2]^{1/2} = .754 \\
 \theta &= \tan^{-1}\left(\frac{.183}{-.731}\right) = 194.055^\circ \\
 &\quad \text{or } -165.945^\circ
 \end{aligned}$$

$$W = \underline{0.754 \angle 194.055^\circ} \text{ (polar form),}$$

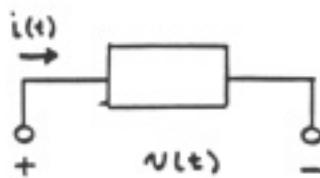
$$W = \underline{-0.731 + j.183} \text{ (rectangular form)}$$

1.c. (5 pts) Use phasors to find the quantities  $A$  and  $\theta$  defined by

$$\begin{aligned}\cos(10t + \pi/4) - \sin(10t + \pi/6) &= A \cos(10t + \theta) \\&= \cos(10t + \pi/4) + \cos(10t + \frac{\pi}{6} + \frac{\pi}{2}) \\&= e^{j\pi/4} + e^{j2\pi/3} \\&= \frac{1}{\sqrt{2}} (1+j) + \left(-\frac{1}{2} + j\frac{\sqrt{3}}{2}\right) \\&= 0.207 + j1.573 \\&\therefore r = [(0.207)^2 + (1.573)^2]^{1/2} = 1.587 \\&\theta = \tan^{-1} \left[ \frac{1.573}{0.207} \right] = 1.44 \text{ rad} \\&\quad (82.5^\circ)\end{aligned}$$

$$A = \underline{1.587}, \quad \theta = \underline{82.5^\circ (1.44 \text{ rad})}$$

1.d. (5 pts) Consider the device below:



where the voltage  $v(t)$  and the current  $i(t)$  are

$$\begin{aligned}v(t) &= 2 \cos(\omega t + \pi/3) \\i(t) &= \cos(\omega t + \pi/4)\end{aligned}$$

Find the impedance  $Z$  of the device and the average power  $P$  dissipated by the device for  $\omega = 20$  rad/sec.

$$V = 2 \angle \pi/3$$

$$I = 1 \angle \pi/4$$

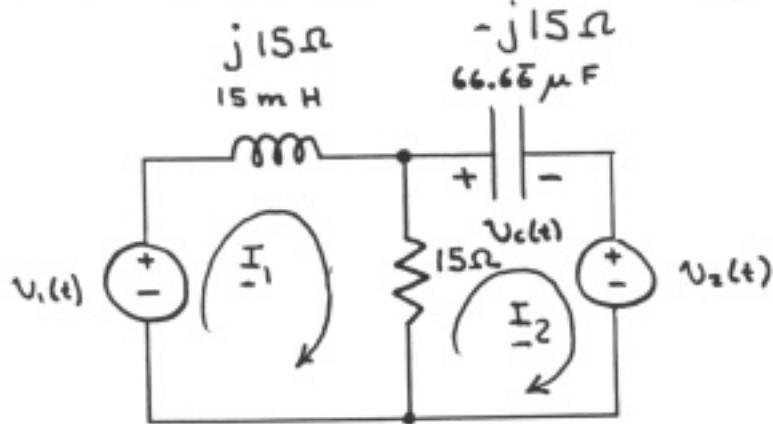
$$Z = \frac{V}{I} = 2 \angle \pi/12 = 1.932 + j0.518 \Omega$$

$$P = \frac{1}{2} \operatorname{Re}[V \cdot I^*] = \frac{1}{2} \operatorname{Re}[2 \angle \pi/12] = 0.966 W$$

$$Z = \underline{1.932 + j0.518 \Omega}, \quad P = \underline{0.966 W}$$

**Problem: 2 (20 pts)**

Consider the following circuit where  $v_1(t) = 30 \cos(1000t)$  V and  $v_2(t) = 30 \cos(1000t - 90^\circ)$  V:



2.a (5 pts): Draw and label the mesh currents on the circuit diagram.

2.b (10pts) Write down the mesh equations.

$$\begin{aligned} \text{mesh 1} \quad & -30 + j15\bar{I}_1 + 15(\bar{I}_1 - \bar{I}_2) = 0 \\ & 15(\bar{I}_2 - \bar{I}_1) - j15\bar{I}_2 - 30j = 0 \\ \begin{bmatrix} 15(1+j) & -15 \\ -15 & 15(1-j) \end{bmatrix} \begin{bmatrix} \bar{I}_1 \\ \bar{I}_2 \end{bmatrix} &= \begin{bmatrix} 30 \\ 30j \end{bmatrix} \end{aligned}$$

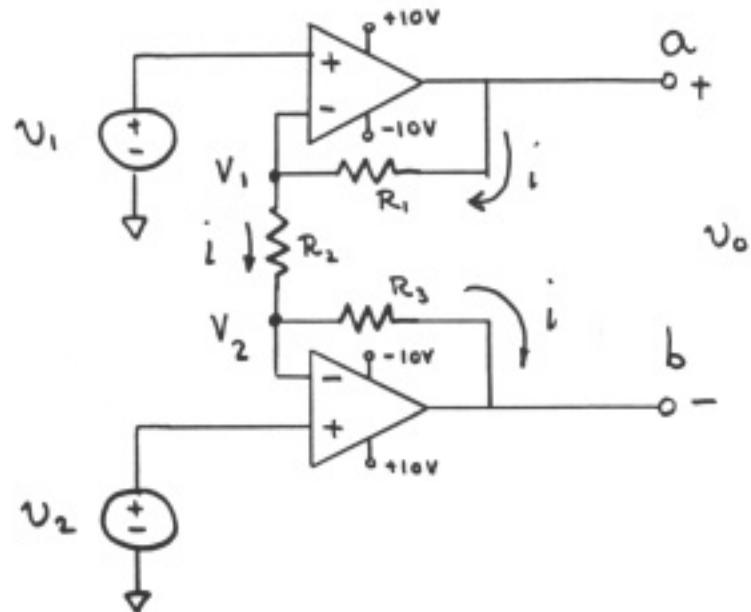
2.c (5 pts) Solve the mesh equations and find  $v_c(t)$ .

$$\begin{aligned} \therefore (1+j)\bar{I}_1 - \bar{I}_2 &= 2 \\ \bar{I}_1 - (1-j)\bar{I}_2 &= -2j \\ \therefore \bar{I}_1 &= 2A \quad \bar{I}_2 = 2j \text{ A} \\ V_c &= (-j15)(\bar{I}_2) = 30V \end{aligned}$$

$$v_c(t) = \underline{30 \cos(1000t)} \text{ V}$$

Problem: 3 (20 pts)

Consider the following circuit:



3.a (10 pts) Assume that Op-Amp is ideal and not in saturation. Find the output voltage  $v_o$  in terms of the input voltages  $v_1$ ,  $v_2$ , and resistances  $R_1$ ,  $R_2$  and  $R_3$ .

$$\text{By Golden Rule #1 } i = (V_1 - V_2) / R_2$$

$$\text{By Golden Rule #2 } i = V_o / (R_1 + R_2 + R_3)$$

$$\therefore \frac{V_1 - V_2}{R_2} = \frac{V_o}{R_1 + R_2 + R_3} \implies V_o = \frac{R_1 + R_2 + R_3}{R_2} (V_1 - V_2)$$

$$v_o = \frac{R_1 + R_2 + R_3}{R_2} (V_1 - V_2)$$

3.b (10 pts) If  $v_2 = 0$  and if  $R_1 = R_2 = R_3 = 1000\Omega$  determine how large  $v_1$  can be before the top Op-Amp saturates.

$$\text{Note: } V_2 = 0. \text{ Thus by a voltage divider } V_1 = \frac{R_2}{R_1 + R_2} V_a$$

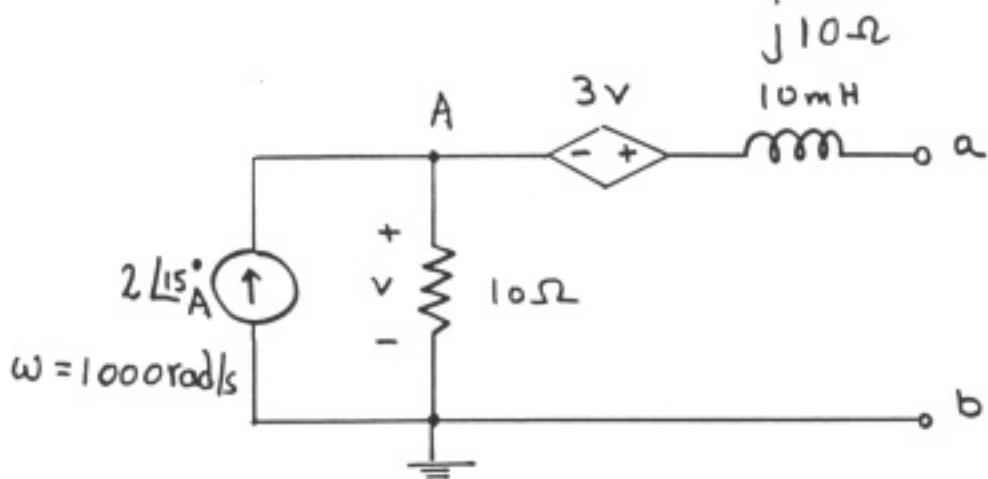
Saturation occurs for  $V_a > 10V$

$$\therefore \max V_1 = 5V$$

$$\max v_1 = 5V$$

Problem: 4 (20 pts)

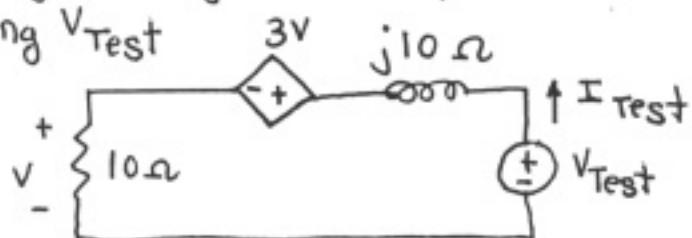
Consider the following circuit:



4.a (15 pts) Find the Thevenin equivalent voltage  $V_{th}$  and impedance  $Z_{th}$ .

$$V_T = V_{OC} = V_A + 3V = 4V_A = (4 \times 2 \angle 15^\circ) \times 10 = 80 \angle 15^\circ \text{ V}$$

Find  $Z_T$  by turning off independent current source & applying  $V_{Test}$



$$V_{th} = 80 \angle 15^\circ, \quad Z_{th} = 40 + j10\Omega$$

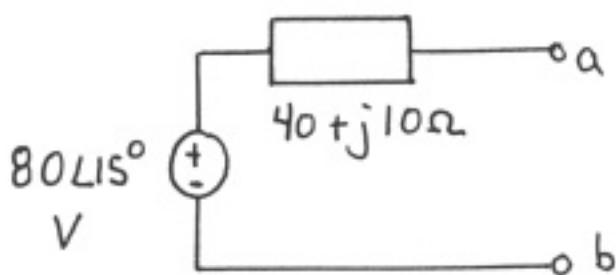
4.b (5 pts) Draw the Thevenin equivalent circuit.

$$\text{Apply KVL} \quad \therefore -V - 3V - (j10)I_{Test} + V_{Test} = 0$$

$$V = 10 I_{Test}$$

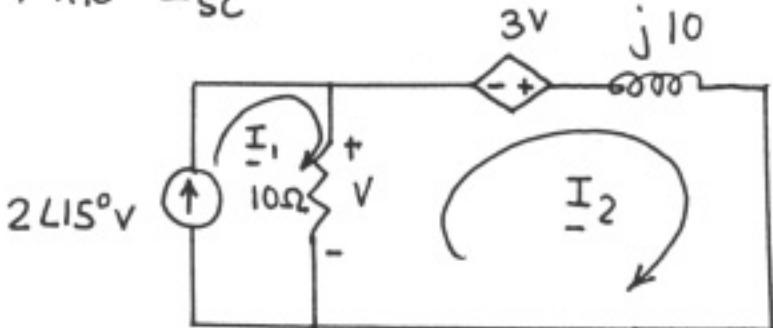
$$\therefore -40I_{Test} - j10I_{Test} + V_{Test} = 0$$

$$Z_{th} = \frac{V_{Test}}{I_{Test}} = 40 + j10\Omega$$



Alternate method for finding  $Z_T$ .

Find  $I_{sc}$



$$\text{mesh 1} \quad I_1 = 2 \angle 15^\circ$$

$$\text{mesh 2} \quad (I_2 - I_1) 10 - 3V + j10 I_2 = 0$$

$$V = 10(I_1 - I_2)$$

$$\therefore 10(I_2 - I_1) - 30(I_1 - I_2) + j10 I_2 = 0$$

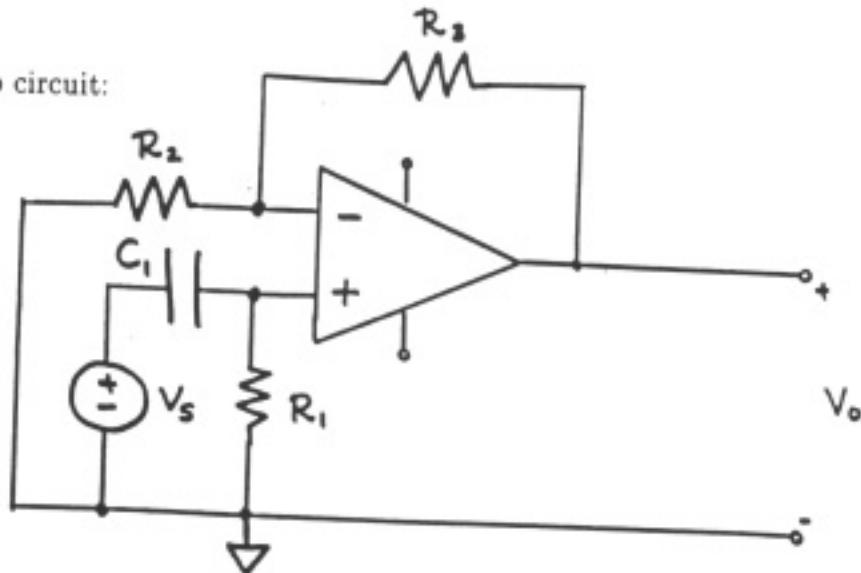
substitute  $I_1 = 2 \angle 15^\circ$  in above equation to get

$$I_{sc} = I_2 = \frac{80 \angle 15^\circ}{40 + j10}$$

$$Z_T = \frac{V_{oc}}{I_{sc}} = \frac{80 \angle 15^\circ}{\frac{80 \angle 15^\circ}{40 + j10}} = 40 + j10 \Omega$$

Problem: 5 (20 pts)

Consider the following Op-Amp circuit:



Assume that the Op-Amp is ideal and not in saturation. Compute the transfer function  $H(j\omega)$  for  $v_o(t)/v_s(t)$

$$V_{(+)} = \frac{R_1}{R_1 + \frac{1}{j\omega C_1}} V_s = \frac{j\omega R_1 C_1}{1 + j\omega R_1 C_1} V_s \quad (\text{Voltage divider})$$

KCL @ (-) terminal

$$V_{(-)} = V_{(+)}$$

(Golden Rule #1)

$$\frac{0 - V_{(-)}}{R_2} + \frac{V_o - V_{(-)}}{R_3} = 0$$

$$\therefore V_o = \left(1 + \frac{R_3}{R_2}\right) V_{(-)} = \left(1 + \frac{R_3}{R_2}\right) \frac{j\omega R_1 C_1}{1 + j\omega R_1 C_1}$$

$$H(j\omega) = \frac{\left(1 + \frac{R_3}{R_2}\right) \frac{j\omega R_1 C_1}{1 + j\omega R_1 C_1}}{1 + j\omega R_1 C_1}$$