1. Problem 7.55 of text

a) For n repeaters in cascade, the probability of i out of n repeaters to produce an error is given by the binomial distribution

$$P_i = \left(\begin{array}{c}n\\i\end{array}\right) p^i (1-p)^{n-i}$$

However, there is a bit error at the output of the terminal receiver only when an odd number of repeaters produces an error. Hence, the overall probability of error is

$$P_n = P_{\text{odd}} = \sum_{i=\text{odd}} \binom{n}{i} p^i (1-p)^{n-i}$$

Let  $P_{\text{even}}$  be the probability that an even number of repeaters produces an error. Then

$$P_{\text{even}} = \sum_{i=\text{even}} \begin{pmatrix} n\\i \end{pmatrix} p^i (1-p)^{n-i}$$

and therefore,

$$P_{\text{even}} + P_{\text{odd}} = \sum_{i=0}^{n} \binom{n}{i} p^{i} (1-p)^{n-i} = (p+1-p)^{n} = 1$$

One more relation between  $P_{\text{even}}$  and  $P_{\text{odd}}$  can be provided if we consider the difference  $P_{\text{even}} - P_{\text{odd}}$ . Clearly,

$$P_{\text{even}} - P_{\text{odd}} = \sum_{i=\text{even}} \binom{n}{i} p^i (1-p)^{n-i} - \sum_{i=\text{odd}} \binom{n}{i} p^i (1-p)^{n-i}$$
  
$$\stackrel{a}{=} \sum_{i=\text{even}} \binom{n}{i} (-p)^i (1-p)^{n-i} + \sum_{i=\text{odd}} \binom{n}{i} (-p)^i (1-p)^{n-i}$$
  
$$= (1-p-p)^n = (1-2p)^n$$

where the equality (a) follows from the fact that  $(-1)^i$  is 1 for *i* even and -1 when *i* is odd. Solving the system

$$P_{\text{even}} + P_{\text{odd}} = 1$$
  
 $P_{\text{even}} - P_{\text{odd}} = (1 - 2p)^n$ 

we obtain

$$P_n = P_{\text{odd}} = \frac{1}{2}(1 - (1 - 2p)^n)$$

**b**) Expanding the quantity  $(1-2p)^n$ , we obtain

$$(1-2p)^n = 1-n2p + \frac{n(n-1)}{2}(2p)^2 + \cdots$$

Since,  $p \ll 1$  we can ignore all the powers of p which are greater than one. Hence,

$$P_n \approx \frac{1}{2}(1 - 1 + n2p) = np = 100 \times 10^{-6} = 10^{-4}$$

## 2. Problem 7.56 of text

The overall probability of error is approximated by

$$P(e) = KQ\left[\sqrt{\frac{\mathcal{E}_b}{N_0}}\right]$$

Thus, with  $P(e) = 10^{-6}$  and K = 100, we obtain the probability of each repeater  $P_r = Q\left[\sqrt{\frac{E_b}{N_0}}\right] = 10^{-8}$ . The argument of the function  $Q[\cdot]$  that provides a value of  $10^{-8}$  is found from tables to be

$$\sqrt{\frac{\mathcal{E}_b}{N_0}} = 5.61$$

Hence, the required  $\frac{\mathcal{E}_b}{N_0}$  is  $5.61^2 = 31.47$ 

3. Problem 7.57 of text

a) The antenna gain for a parabolic antenna of diameter D is

$$G_R = \eta \left(\frac{\pi D}{\lambda}\right)^2$$

If we assume that the efficiency factor is 0.5, then with

$$\lambda = \frac{c}{f} = \frac{3 \times 10^8}{10^9} = 0.3 \text{ m}$$
  $D = 3 \times 0.3048 \text{ m}$ 

we obtain

$$G_R = G_T = 45.8458 = 16.61 \text{ dB}$$

**b**) The effective radiated power is

$$EIRP = P_T G_T = G_T = 16.61 \text{ dB}$$

c) The received power is

$$P_R = \frac{P_T G_T G_R}{\left(\frac{4\pi d}{\lambda}\right)^2} = 2.995 \times 10^{-9} = -85.23 \text{ dB} = -55.23 \text{ dBm}$$

Note that

dBm = 
$$10\log_{10}\left(\frac{\text{actual power in Watts}}{10^{-3}}\right) = 30 + 10\log_{10}(\text{power in Watts})$$

- 4. Problem 7.58 of text
  - **a**) The antenna gain for a parabolic antenna of diameter *D* is

$$G_R = \eta \left(\frac{\pi D}{\lambda}\right)^2$$

If we assume that the efficiency factor is 0.5, then with

$$\lambda = \frac{c}{f} = \frac{3 \times 10^8}{10^9} = 0.3 \text{ m}$$
 and  $D = 1 \text{ m}$ 

we obtain

$$G_R = G_T = 54.83 = 17.39 \text{ dB}$$

**b**) The effective radiated power is

$$EIRP = P_T G_T = 0.1 \times 54.83 = 7.39 \text{ dB}$$

**c**) The received power is

$$P_R = \frac{P_T G_T G_R}{\left(\frac{4\pi d}{\lambda}\right)^2} = 1.904 \times 10^{-10} = -97.20 \text{ dB} = -67.20 \text{ dBm}$$

5. Problem 7.59 of text

The wavelength of the transmitted signal is

$$\lambda = \frac{3 \times 10^8}{10 \times 10^9} = 0.03 \text{ m}$$

The gain of the parabolic antenna is

$$G_R = \eta \left(\frac{\pi D}{\lambda}\right)^2 = 0.6 \left(\frac{\pi 10}{0.03}\right)^2 = 6.58 \times 10^5 = 58.18 \text{ dB}$$

The received power at the output of the receiver antenna is

$$P_R = \frac{P_T G_T G_R}{(4\pi \frac{d}{\lambda})^2} = \frac{3 \times 10^{1.5} \times 6.58 \times 10^5}{(4 \times 3.14159 \times \frac{4 \times 10^7}{0.03})^2} = 2.22 \times 10^{-13} = -126.53 \text{ dB}$$

6. Problem 7.60 of text

**a**) Since  $T = 300^0 K$ , it follows that

$$N_0 = kT = 1.38 \times 10^{-23} \times 300 = 4.14 \times 10^{-21}$$
 W/Hz

If we assume that the receiving antenna has an efficiency  $\eta = 0.5$ , then its gain is given by

$$G_R = \eta \left(\frac{\pi D}{\lambda}\right)^2 = 0.5 \left(\frac{3.14159 \times 50}{\frac{3 \times 10^8}{2 \times 10^9}}\right)^2 = 5.483 \times 10^5 = 57.39 \text{ dB}$$

Hence, the received power level is

$$P_R = \frac{P_T G_T G_R}{(4\pi \frac{d}{\lambda})^2} = \frac{10 \times 10 \times 5.483 \times 10^5}{(4 \times 3.14159 \times \frac{10^8}{0.15})^2} = 7.8125 \times 10^{-13} = -121.07 \text{ dB}$$

**b**) If  $\frac{\mathcal{E}_b}{N_0} = 10 \text{ dB} = 10$ , then

$$R = \frac{P_R}{N_0} \left(\frac{\mathcal{E}_b}{N_0}\right)^{-1} = \frac{7.8125 \times 10^{-13}}{4.14 \times 10^{-21}} \times 10^{-1} = 1.8871 \times 10^7 = 18.871 \text{ Mbits/sec}$$

7. Problem 7.62 of text

The wavelength of the transmission is

$$\lambda = \frac{c}{f} = \frac{3 \times 10^8}{4 \times 10^9} = 0.75 \text{ m}$$

If 1 MHz is the passband bandwidth, then the rate of binary transmission is  $R_b = W = 10^6$  bps. Hence, with  $N_0 = 4.1 \times 10^{-21}$  W/Hz we obtain

$$\frac{P_R}{N_0} = R_b \frac{\mathcal{E}_b}{N_0} \Longrightarrow 10^6 \times 4.1 \times 10^{-21} \times 10^{1.5} = 1.2965 \times 10^{-13}$$

The transmitted power is related to the received power through the relation

$$P_R = \frac{P_T G_T G_R}{(4\pi \frac{d}{\lambda})^2} \Longrightarrow P_T = \frac{P_R}{G_T G_R} \left(4\pi \frac{d}{\lambda}\right)^2$$

Substituting in this expression the values  $G_T = 10^{0.6}$ ,  $G_R = 10^5$ ,  $d = 36 \times 10^6$  and  $\lambda = 0.75$  we obtain

$$P_T = 0.1185 = -9.26 \text{ dBW}$$

Alternate Solution. Since it mentions reliable communication we assume that the data rate is the "capacity achieving" data rate. So (see lecture notes 1)

$$R/W = \log 2(1 + \frac{E_b}{N_0}\frac{R}{W})$$

Since  $E_b/N_0 = 15$  dB or  $E_b/N_0 = 31.62$  This correspondes to R/W = 8. So R=8Mbps. Now using R = 8 Mbps and  $N_0 = 4.1 \times 10^{-21}$  we get

$$\frac{P_r/R}{N_0} = 31.62$$

$$P_r = 31.62(RN_0)$$

$$P_r = 31.62(32 \times 10^{-15})$$

$$P_r = 1.01 \times 10^{-12}$$

$$P_T = \frac{P_R}{G_T G_R} \left(4\pi \frac{d}{\lambda}\right)^2$$
$$= \frac{10^{-12}}{3} \left(4\pi \frac{36 \times 10^6}{.75}\right)^2$$
$$= 117 dB = 500 GW atts$$

This is obviously not a possible solution. However, from the problem statement this is a reasonable approach to take to solve the problem. However modulation techniques transmitting 8Mbps in a 1 MHz bandwidth would require a very linear (and thus very inefficient) amplifier. So the restriction that the communication use BPSK with data rate =bandwidth is resonable.

8. Problem 7.63 of text

Since  $T = 290^{\circ} + 15^{\circ} = 305^{\circ}K$ , it follows that

$$N_0 = kT = 1.38 \times 10^{-23} \times 305 = 4.21 \times 10^{-21}$$
 W/Hz

The transmitting wavelength  $\lambda$  is

$$\lambda = \frac{c}{f} = \frac{3 \times 10^8}{2.3 \times 10^9} = 0.130 \text{ m}$$

Hence, the gain of the receiving antenna is

$$G_R = \eta \left(\frac{\pi D}{\lambda}\right)^2 = 0.55 \left(\frac{3.14159 \times 64}{0.130}\right)^2 = 1.3156 \times 10^6 = 61.19 \text{ dB}$$

and therefore, the received power level is

$$P_R = \frac{P_T G_T G_R}{(4\pi_{\overline{\lambda}}^d)^2} = \frac{17 \times 10^{2.7} \times 1.3156 \times 10^6}{(4 \times 3.14159 \times \frac{1.6 \times 10^{11}}{0.130})^2} = 4.686 \times 10^{-12} = -113.29 \text{ dB}$$

If  $\mathcal{E}_b/N_0 = 6 \text{ dB} = 10^{0.6}$ , then

$$R = \frac{P_R}{N_0} \left(\frac{\mathcal{E}_b}{N_0}\right)^{-1} = \frac{4.686 \times 10^{-12}}{4.21 \times 10^{-21}} \times 10^{-0.6} = 4.4312 \times 10^9 = 4.4312 \text{ Gbits/sec}$$