

## Lecture 7

Goals:

- Signals as Vectors, Noise as Vectors
- Optimum Detection in AWGN

VII-1

## Decomposition of Signal and Noise

Given a set of signals  $s_0(t), \dots, s_{M-1}(t)$  there exists a set of orthonormal signals  $\phi_0(t), \phi_1(t), \dots, \phi_{N-1}(t)$  with  $N \leq M$  such that

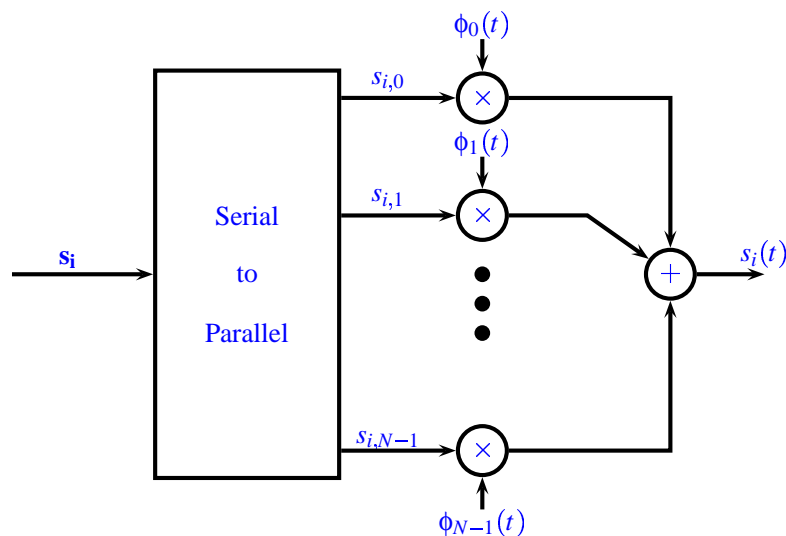
$$s_i(t) = \sum_{m=0}^{N-1} s_{i,m} \phi_m(t)$$

For any complete orthonormal set of signals  $\phi_0(t), \phi_1(t), \dots$  we can represent a noise process as random variables and deterministic orthonormal functions

$$n(t) = \sum_{m=0}^{\infty} n_m \phi_m(t)$$

VII-2

## Signal



VII-3

## Decomposition of Signal and Noise

Consider a communication system that transmits one of  $M$  signals.  $s_0(t), \dots, s_{M-1}(t)$  in additive white Gaussian noise. s Then given  $s_i(t)$  was transmitted the received signal is

$$\begin{aligned} r(t) &= s_i(t) + n(t) \\ &= \sum_{m=0}^{\infty} (s_{i,m} + n_m) \phi_m(t) \end{aligned}$$

Define  $r_m = s_{i,m} + n_m$ . Then

$$r(t) = \sum_{m=0}^{\infty} r_m \phi_m(t)$$

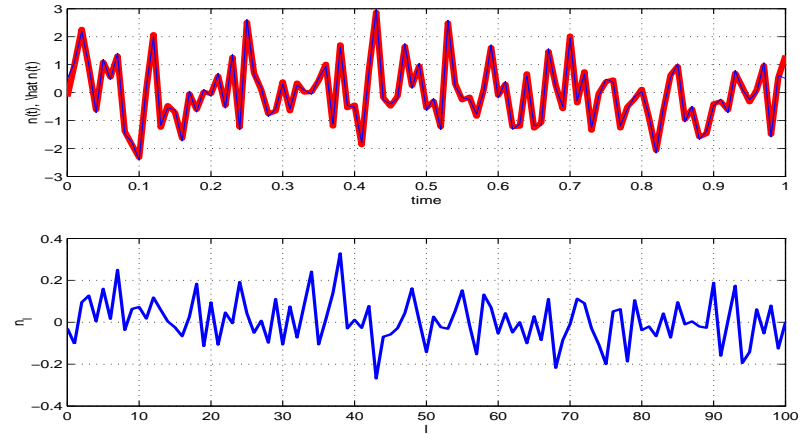
VII-4

We can determine the (random) variable  $r_m$  by

$$r_m = \int r(t) \phi_m^*(t) dt$$

VII-5

## Decomposition of Noise



VII-6

## Example

$$s_0(t) = A p_T(t)$$

$$s_1(t) = -A p_T(t)$$

Let  $\phi_l(t) = \sqrt{\frac{1}{T}} \exp\{j2\pi l f_0 t\} p_T(t)$  where  $f_0 = 1/T$ . Then

$$s_0(t) = \sqrt{E} \phi_0(t)$$

$$s_1(t) = -\sqrt{E} \phi_0(t)$$

$$n(t) = \sum_{m=0}^{\infty} n_m \phi_m(t)$$

$$r(t) = \sum_{m=0}^{\infty} r_m \phi_m(t)$$

VII-7

$$\begin{aligned} r_m &= \int r(t) \phi_m^*(t) dt \\ &= \int (s_i(t) + n(t)) \phi_m^*(t) dt \\ &= s_{i,m} + n_m \end{aligned}$$

Note that we can recover completely  $r(t)$  if we know the coefficients  $r_m, m = 0, 1, \dots$ . So the optimal decision based on observing  $r_0, r_1, \dots$  is also the optimal decision based on observing  $r(t)$ . Given signal  $s_i(t)$  is transmitted we can determine the probability density of  $r_m$  as follows. First,  $r_m$  is Gaussian since it is the result of integrating Gaussian noise. Second the mean of  $r_m$  is  $s_{i,m}$  and the variance is  $N_0/2$ . So the probability density of  $r_m$  conditioned on signal  $i$  transmitted (event  $H_i$ ) is

$$\begin{aligned} p_i(r_m) &= f_{r_m|H_i}(r_m) \\ &= \frac{1}{\sqrt{2\pi} \sqrt{N_0/2}} \exp\left\{-\frac{(r_m - s_{i,m})^2}{2(N_0/2)}\right\} \end{aligned}$$

VII-8

Next note that  $r_m$  is independent of  $r_m$  for  $m \neq n$ . Thus

$$\begin{aligned} f_{r_0, r_1, \dots, r_k | H_i}(x_0, x_1, x_2, \dots, x_k) &= \prod_{m=0}^k f_{r_m | H_i}(x_m) \\ &= \prod_{m=0}^k p_i(x_m) \end{aligned}$$

VII-9

## M-ary Detection Problem

Consider the problem of deciding which of  $M$  hypothesis is true based on observing a random variable (vector)  $r$ . The performance criteria we consider is the average error probability. That is the probability of deciding anything except hypothesis  $H_j$  when hypothesis  $H_j$  is true.

The underlying model is that there is a conditional probability density (mass) function of the observation  $r$  given each hypothesis  $H_j$ . There are disjoint decision regions  $R_0, R_1, \dots, R_{M-1}$ . When  $r \in R_m$  the receiver decides  $H_m$ .

$$\begin{aligned} E[P_e] &= \sum_{i=0}^{M-1} P_{e,i} \pi_i = \sum_{i=0}^{M-1} P\{\text{don't decide } H_i | H_i\} \pi_i \\ &= \sum_{i=0}^{M-1} P\{r \in \bigcup_{l=0, l \neq i}^{M-1} R_l | H_i\} \pi_i \end{aligned}$$

VII-10

$$\begin{aligned} &= \sum_{i=0}^{M-1} \left[ \sum_{l \neq i} P\{\text{decide } H_l | H_i \text{ true}\} \right] \pi_i \\ &= \sum_{i=0}^{M-1} [1 - P\{\text{decide } H_i | H_i \text{ true}\}] \pi_i \\ &= \sum_{i=0}^{M-1} \pi_i - \sum_{i=0}^{M-1} \int_{R_i} p_i(r) \pi_i dr \\ &= 1 - \sum_{i=0}^{M-1} \int_{R_i} p_i(r) \pi_i dr . \end{aligned}$$

The decision rule that minimizes average probability of error assigns  $r$  to  $R_i$  if  $p_i(r) \pi_i = \max_{0 \leq j \leq M-1} p_j(r) \pi_j$ . Let  $p(r)$  be an arbitrary density function that is nonzero everywhere  $p_i(r)$  is nonzero then an equivalent decision rule is to assign  $r$  to  $R_i$  if

$$\frac{p_i(r)}{p(r)} \pi_i = \max_{0 \leq j \leq M-1} \frac{p_j(r)}{p(r)} \pi_j .$$

VII-11

Thus for  $M$  hypotheses the decision rule that minimizes average error probability is to choose  $i$  so that  $p_i(r) \pi_i > p_j(r) \pi_j$ ,  $\forall j \neq i$ . Let

$$\Lambda_{i,j} = \frac{p_i(r)}{p_j(r)}$$

where  $i = 0, 1, \dots, M-1$ ,  $j = 0, 1, \dots, M-1$ . Then the optimal decision rule is:

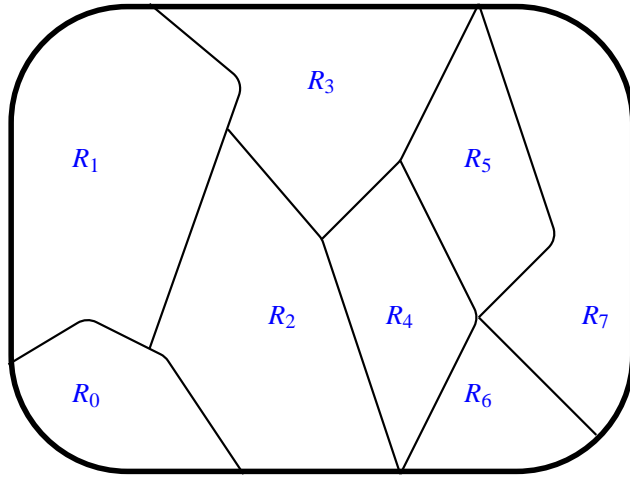
$$\text{Choose } i \text{ if } \Lambda_{i,j} > \frac{\pi_j}{\pi_i} \text{ for all } j \neq i.$$

We will usually assume  $\pi_i = \frac{1}{M} \forall i$ . (If not we should do source encoding to reduce the entropy (rate)). For this case the optimal decision rule is

$$\text{Choose } i \text{ if } \Lambda_{i,j} > 1 \quad \forall j \neq i.$$

VII-12

## Decision Regions



VII-13

## Example 1: Additive White Gaussian Noise

Consider three signals in additive white Gaussian noise. For additive white Gaussian noise  $K(s, t) = \frac{N_0}{2} \delta(t - s)$ . Let  $\{\phi_i(t)\}_{i=0}^{\infty}$  be any complete orthonormal set on  $[0, T]$ . Consider the case of 3 signals. Find the decision rule to minimize average error probability. First expand the noise using orthonormal set of functions and random variables.

$$n(t) = \sum_{i=0}^{\infty} n_i \phi_i(t)$$

where  $E[n_i] = 0$  and  $\text{Var}[n_i] = N_0/2$  and  $\{n_i\}_{i=0}^{\infty}$  is an independent identically distributed (i.i.d.) sequence of random variables with Gaussian density functions.

Let

$$s_0(t) = \phi_0(t) + 2\phi_1(t)$$

VII-14

$$s_1(t) = 2\phi_0(t) + \phi_1(t)$$

$$s_2(t) = \phi_0(t) - 2\phi_1(t)$$

Note that the energy of each of the three signals is the same, i.e.  $\int_0^T s_i^2(t) dt = \|s_i\|^2 = 5$ . Then we have a three hypothesis testing problem.

$$H_0 : r(t) = s_0(t) + n(t) = \sum_{i=0}^{\infty} (s_{0,i} + n_i) \phi_i(t)$$

$$H_1 : r(t) = s_1(t) + n(t) = \sum_{i=0}^{\infty} (s_{1,i} + n_i) \phi_i(t)$$

$$H_2 : r(t) = s_2(t) + n(t) = \sum_{i=0}^{\infty} (s_{2,i} + n_i) \phi_i(t)$$

The decision rule to minimize the average error probability is given as follows

$$\text{Decide } H_i \text{ if } \pi_i p_i(\mathbf{r}) = \max_j \pi_j p_j(\mathbf{r})$$

First let us consider the first  $L+1$  variables and normalize each side by the

VII-15

density function for the noise alone. The noise density function for  $L+1$  variables is

$$p^{(L)}(\mathbf{r}) = \left( \frac{1}{\sqrt{2\pi N_0/2}} \right)^L \exp \left\{ -\frac{1}{2 \frac{N_0}{2}} \sum_{m=0}^L r_m^2 \right\}$$

The the optimal decision rule is equivalent to

$$\text{Decide } H_i \text{ if } \pi_i \frac{p_i(\mathbf{r})}{p(\mathbf{r})} = \max_j \pi_j \frac{p_j(\mathbf{r})}{p(\mathbf{r})}.$$

As usual assume  $\pi_i = 1/M$ . Then

$$\begin{aligned} \frac{p_0^{(L)}(\mathbf{r})}{p^{(L)}(\mathbf{r})} &= \frac{\left( \frac{1}{\sqrt{2\pi N_0/2}} \right)^L \exp \left\{ -\frac{1}{2 \frac{N_0}{2}} [\sum_{i=0,1} (r_i - s_{0,i})^2 + \sum_{i=2}^L r_i^2] \right\}}{\left( \frac{1}{\sqrt{2\pi N_0/2}} \right)^L \exp \left\{ -\frac{1}{2 \frac{N_0}{2}} \sum_{i=0}^L r_i^2 + \sum_{i=2}^L r_i^2 \right\}} \\ &= \exp \left\{ -\frac{1}{N_0} \left[ \sum_{i=0,1} (r_i - s_{0,i})^2 - r_i^2 \right] \right\} \end{aligned}$$

VII-16

$$= \exp\left\{+\frac{1}{N_0}[2r_1 + 4r_2 - 5]\right\}.$$

Now since the above doesn't depend on  $L$  we can let  $L \rightarrow \infty$  and the result is the same, i.e.

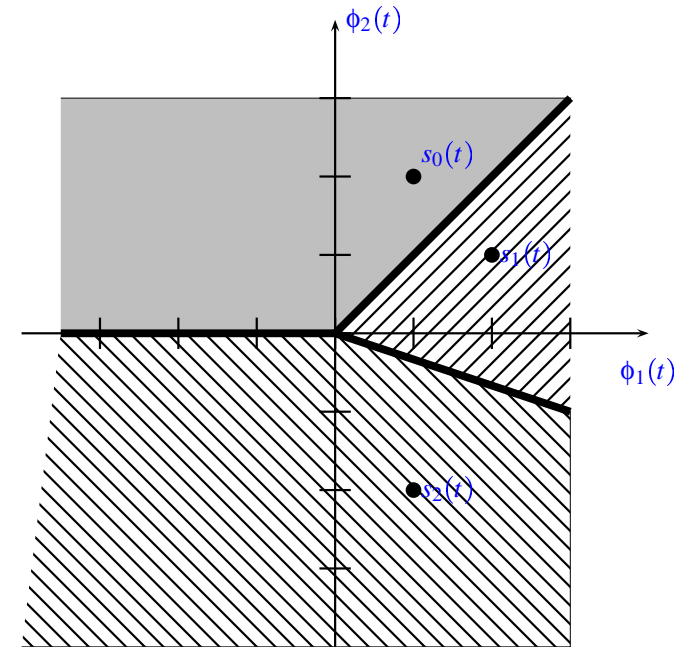
$$\frac{p_0(\mathbf{r})}{p(\mathbf{r})} \triangleq \lim_{L \rightarrow \infty} \frac{p_0^{(L)}(\mathbf{r})}{p^{(L)}(\mathbf{r})} = \exp\left\{+\frac{1}{N_0}[2r_0 + 4r_1 - 5]\right\}.$$

Similarly

$$\frac{p_1(\mathbf{r})}{p(\mathbf{r})} = \exp\left\{+\frac{1}{N_0}[4r_0 + 2r_1 - 5]\right\}$$

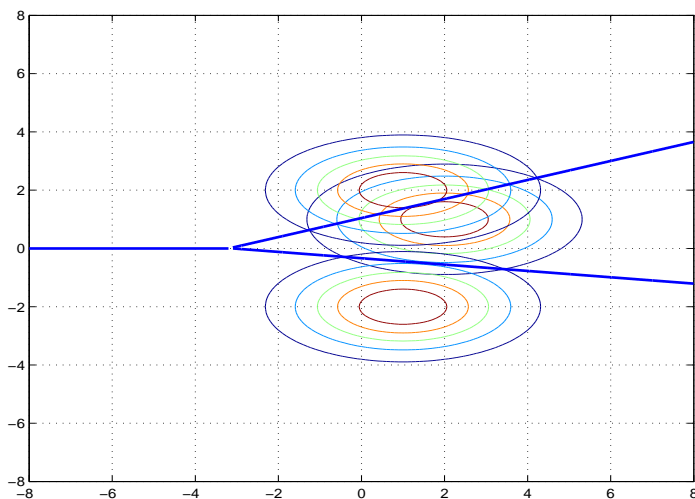
$$\frac{p_2(\mathbf{r})}{p(\mathbf{r})} = \exp\left\{+\frac{1}{N_0}[2r_0 - 4r_1 - 5]\right\}.$$

VII-17



VII-18

## Decision Regions



VII-19

## Likelihood Ratio for Real Signals in AWGN

Assume two signals in Gaussian noise.

$$H_0 : r(t) = s_0(t) + n(t)$$

$$H_1 : r(t) = s_1(t) + n(t)$$

**Goal:** Find decision rule to minimize the average error probability.

Let  $n(t)$  autocorrelation function  $R((s,t) = \frac{N_0}{2}\delta(t-s)$  We assume that  $n(t)$  is a zero mean white Gaussian noise random process.

VII-20

## Karhunen-Loeve Expansion

By Karhunen-Loeve expansion

$$n(t) = \sum_{m=0}^{\infty} n_m \phi_m(t)$$

where  $n_i$  are Gaussian random variables with mean 0 variance  $\frac{N_0}{2}$  and  $E[n_m n_k] = 0, m \neq k$ . Thus  $n_m$  and  $n_k$  are independent. Since  $\{\phi_m(t); m = 0, 1, \dots\}$  is a complete orthonormal set and we assume  $s_j(t)$  has finite energy we have

$$s_j(t) = \sum_{m=0}^{\infty} s_{j,m} \phi_m(t) = \sum_{m=0}^{N-1} s_{j,m} \phi_m(t).$$

This last equality is because we only need a finite ( $N \leq M$ ) orthonormal waveforms to represent a set of  $M$  signals. Equivalently  $s_{j,i} = 0$  for  $i \geq N$ .

VII-21

Thus

$$H_j : r(t) = \sum_{m=0}^{\infty} (s_{j,m} + n_m) \phi_m(t)$$

$$r_m = s_{j,m} + n_m, \quad m = 0, 1, 2, \dots$$

Define

$$\Lambda_{j,i}(L) \triangleq \frac{p_j(r_0, r_1, \dots, r_L)}{p_i(r_0, r_1, \dots, r_L)}.$$

$$\Lambda_{j,i}(r(t)) \triangleq \lim_{L \rightarrow \infty} \Lambda_{j,i}(L)$$

where  $r_m$  is Gaussian with mean  $s_{j,m}$  variance  $N_0/2$ .

$$p_j(r_m) = \frac{1}{\sqrt{N_0\pi}} \exp \left\{ -\frac{1}{N_0} (r_m - s_{j,m})^2 \right\}$$

$$p_j(\underline{r}) = \prod_{m=0}^L p_j(r_m) = \prod_{m=0}^L (\sqrt{N_0\pi})^{-1} \exp \left\{ -\frac{1}{N_0} \sum_{m=0}^L (r_m - s_{j,m})^2 \right\}$$

VII-22

$$\Lambda_{j,l}(L) = \frac{p_j^L(\underline{r})}{p_l^L(\underline{r})} = \frac{\prod_{m=0}^L (\sqrt{N_0\pi})^{-1} \exp \left\{ -\frac{1}{N_0} \sum_{m=0}^L (r_m - s_{j,m})^2 \right\}}{\prod_{m=0}^L (\sqrt{N_0\pi})^{-1} \exp \left\{ -\frac{1}{N_0} \sum_{m=0}^L (r_m - s_{l,m})^2 \right\}}$$

$$= \exp \left\{ -\frac{1}{N_0} \sum_{m=0}^L [r_m^2 - 2r_m s_{j,m} + s_{j,m}^2 - r_m^2 + 2r_m s_{l,m} - s_{l,m}^2] \right\}$$

$$= \exp \left\{ -\frac{1}{N_0} \sum_{m=0}^L [s_{j,m}^2 - s_{l,m}^2 + 2r_m (s_{l,m} - s_{j,m})] \right\}.$$

If we take the limit as  $L \rightarrow \infty$  we get

$$\Lambda_{j,l}(r(t)) = \exp \left\{ -\frac{1}{N_0} (E_0 - E_1 + 2(r, s_l - s_j)) \right\}.$$

$$\Lambda_{j,l}(r(t)) = \exp \left\{ -\frac{1}{N_0} [(s_j, s_j) - (s_l, s_l) + 2(r, s_l) - 2(r, s_j)] \right\}.$$

VII-23

or equivalently

$$\Lambda_{j,l}(r(t)) = \exp \left\{ -\frac{1}{N_0} [\|s_j\|^2 - \|s_l\|^2 + 2(r, s_l - s_j)] \right\}$$

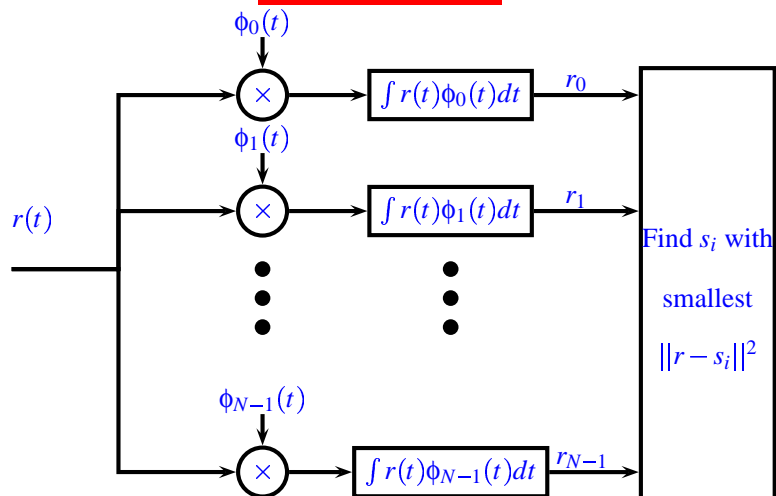
$$= \exp \left\{ -\frac{1}{N_0} [\|r - s_j\|^2 - \|r - s_l\|^2] \right\}$$

The optimum decision rule for additive white Gaussian noise is then to choose  $i$  if

$$\|s_i - r\|^2 = \min_{0 \leq j \leq M-1} \|s_j - r\|^2.$$

VII-24

### Demodulator



VII-25

### Example: $M$ equal energy signals

Now consider the optimum receiver for  $M$ -ary equally likely signals and the associated error probability. Assume the  $M$  signals are equienergy signals and equiprobable. The decision rule derived previously for AWGN in this case simplifies to

$$\text{Decide } H_i \text{ if } \|s_i - r\|^2 = \min_{0 \leq j \leq M-1} \|s_j - r\|^2.$$

Now since the  $M$  signals are equienergy we can write this as

$$\|s_j - r\|^2 = \|s_j\|^2 - 2(s_j, r) + \|r\|^2.$$

The first term above is constant for each  $j$  as is the last term. Thus finding the minimum is equivalent to finding the maximum of

$$(s_j, r).$$

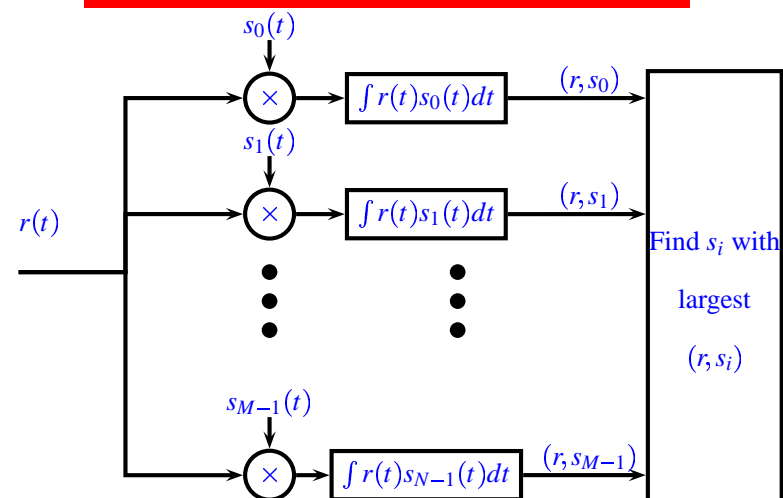
Thus the receiver should compute the inner product between the  $M$  different

VII-26

signals and find the largest such correlation. If the signals are all of duration  $T$ , i.e. zero outside the interval  $[0, T]$  then this is also equivalent to filtering the received signal with a filter with impulse response  $s_j(T - t)$ , sampling the output of the filter at time  $T$  and choosing the largest.

VII-27

### Demodulator (Equal Energy Case)



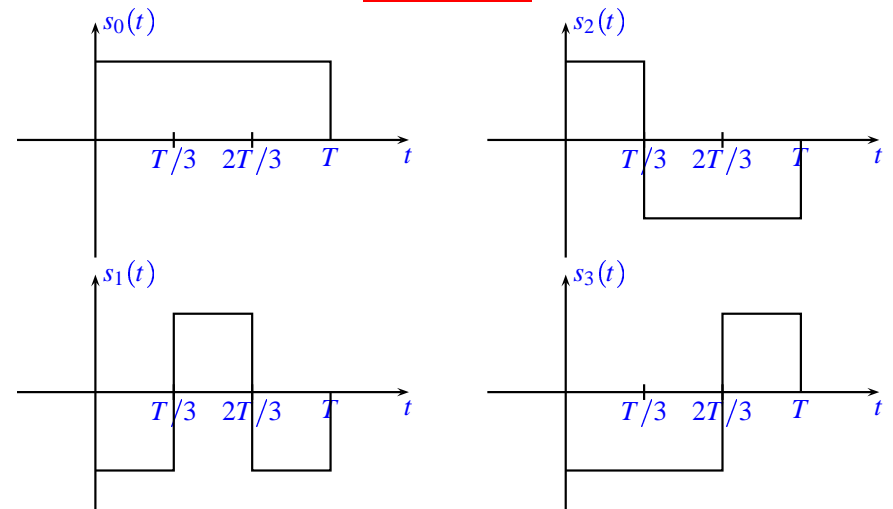
VII-28

## Notes about Optimum Receiver in AWGN

- Consider the case of equally likely signals ( $\pi_0 = \dots = \pi_{M-1} = 1/M$ ).
- The optimum receiver first maps the received signal into a  $N$  dimensional vector. ( $r(t) \rightarrow r$ ).
- The decision region is determined by the perpendicular bisectors of the signal points.
- Then the receiver finds which signal is closest (in Euclidean distance) to the received vector. (Find  $i$  for which  $r \in R_i$ ).

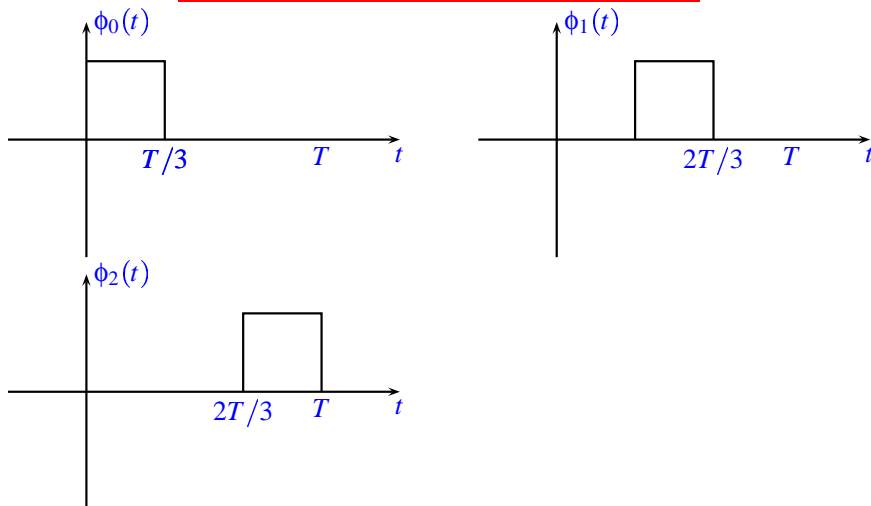
VII-29

## Example



VII-30

## Orthonormal Basis Functions



VII-31

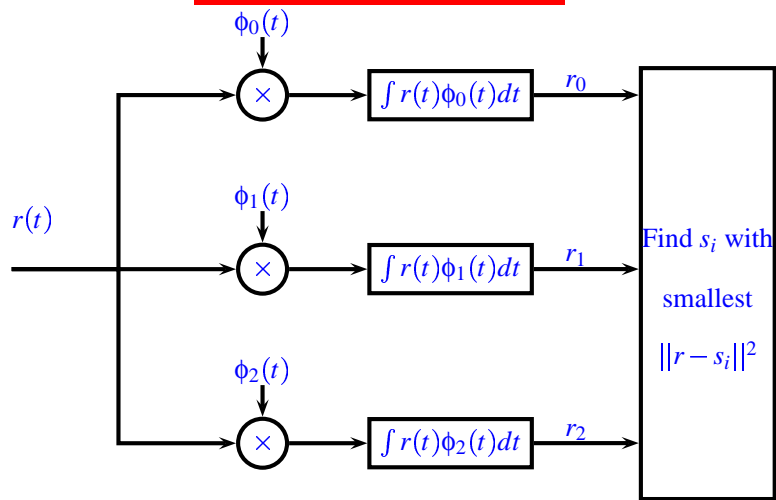
## Signal Vectors

$$\begin{aligned} s_0 &= (+1, +1, +1) \\ s_1 &= (-1, +1, -1) \\ s_2 &= (+1, -1, -1) \\ s_3 &= (-1, -1, +1) \end{aligned}$$

VII-32

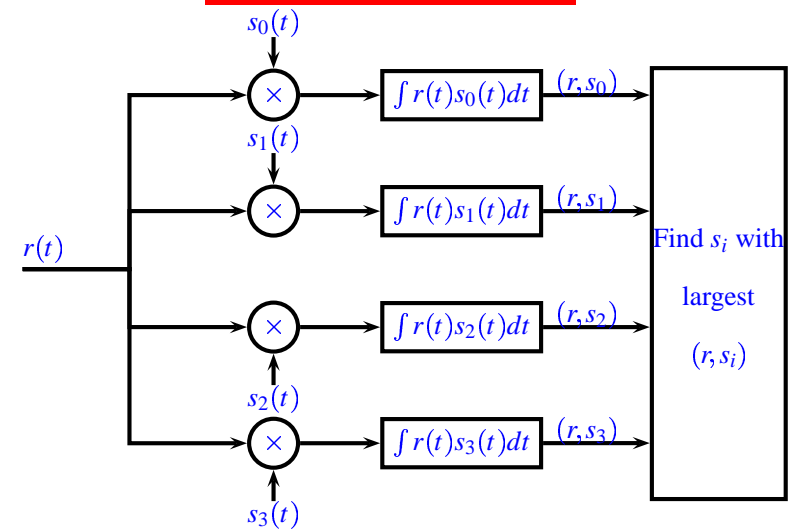


### Optimum Receiver 1



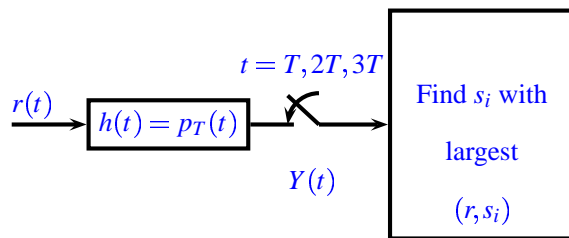
VII-33

### Optimum Receiver 2



VII-34

### Optimum Receiver 3



$$r_0 = Y(T) = \int_0^T r(t) \phi_0(t) dt = \int_0^T r(t) dt$$

$$r_1 = Y(2T) = \int_T^{2T} r(t) \phi_1(t) dt = \int_T^{2T} r(t) dt$$

$$r_2 = Y(3T) = \int_{2T}^{3T} r(t) \phi_2(t) dt = \int_{2T}^{3T} r(t) dt$$

VII-35

### Simplified Calculation

$$\begin{aligned} (r, s_0) &= +r_0 + r_1 + r_2 \\ (r, s_1) &= -r_0 + r_1 - r_2 \\ (r, s_2) &= +r_0 - r_1 - r_2 \\ (r, s_3) &= -r_0 - r_1 + r_2 \end{aligned}$$

First calculated  $x_0, x_1, x_2, x_3$  as follows

$$\begin{aligned} x_0 &= +r_0 \\ x_1 &= -r_0 \\ x_2 &= r_1 + r_2 \\ x_3 &= r_1 - r_2 \end{aligned}$$

VII-36

Then

$$(r, s_0) = x_0 + x_2$$

$$(r, s_1) = x_1 + x_3$$

$$(r, s_2) = x_0 - x_2$$

$$(r, s_3) = x_1 - x_3$$

Thus the calculation requires only 6 additions/subtractions.