

Given: Circuit with: resistors, inductors, capacitors, sinusoidal sources.

Goal: Compute any voltage or current in the circuit. **Know:** also sinusoids.

Want: Amplitude and phase of the sinusoid for each voltage or current.

EX: Voltage source $5 \cos(6t)$ in series with 4Ω resistor and $\frac{1}{2}H$ inductor.

Goal: Compute the current $i(t)$ which flows through all three devices.

Hard KVL $\rightarrow 5 \cos(6t) - 4i - \frac{1}{2} \frac{di}{dt} = 0$. How do we solve this for $i(t)$?

Way: *Trial solution:* $i(t) = A \cos(6t) + B \sin(6t)$. Substitute this in KVL:

$$5 \cos(6t) - 4A \cos(6t) - 4B \sin(6t) + \frac{1}{2} 6A \sin(6t) - \frac{1}{2} 6B \cos(6t) = 0.$$

$$[5 - 4A - 3B] \cos(6t) + [-4B + 3A] \sin(6t) = 0. \text{ Set } t = 0 \text{ and } \frac{\pi}{12} \rightarrow$$

$$5 = 4A + 3B \text{ and } 0 = 3A - 4B. \text{ Solving } \rightarrow A = 0.8 \text{ and } B = 0.6.$$

Soln: $i(t) = 0.8 \cos(6t) + 0.6 \sin(6t) = \cos(6t - 37^\circ)$ using Problem Set #1.

Easy Trial sol'n: $I(t) = Ie^{j6t}$. Substitute in KVL with $5 \cos(6t) \rightarrow 5e^{j6t}$:

Way: $5e^{j6t} - 4Ie^{j6t} - \frac{1}{2} 6jIe^{j6t} = 0 \rightarrow I = 5/(4 + j3) = 1e^{-j37^\circ}$ (e^{j6t} cancels).

$$I(t) = 1e^{j(6t - 37^\circ)} \rightarrow i(t) = \text{Re}[I(t)] = \cos(6t - 37^\circ). \text{ MUCH easier!}$$

Easier: $I = 5/(4 + j\frac{1}{2}) = 5/(4 + j3) = 1e^{-j37^\circ} \rightarrow i(t) = \cos(6t - 37^\circ)$.

Q: What does $5e^{j6t}$ voltage source mean? Complex number voltage?

A: $5e^{j6t} = 5 \cos(6t) + j5 \sin(6t) = 2$ voltage sources connected in series.

Then: Superposition \rightarrow find response to $5 \cos(6t)$ by setting $j5 \sin(6t)$ to 0.

Means: Set $j = 0 \Leftrightarrow$ take the **real part** of response (which has form Ie^{j6t}).

Here: $I(t) = \text{response to } 5e^{j6t} \rightarrow \text{Re}[I(t)] = \text{response to } \text{Re}[5e^{j6t}] = 5 \cos(6t)$.

Also: $\text{Im}[I(t)] = \text{response to } \text{Im}[5e^{j6t}] = 5 \sin(6t)$: Solve 2 problems at once!

Phasors: Represent sinusoid $x(t) = M \cos(\omega t + \theta)$ with complex no. $X = Me^{j\theta}$.

Note: $x(t) = \text{Re}[Xe^{j\omega t}] = \text{Re}[Me^{j\theta} e^{j\omega t}] = \text{Re}[Me^{j(\omega t + \theta)}] = M \cos(\omega t + \theta)$.

EX#1: Simplify $x(t) = 3 \cos(\omega t) + 3 \cos(\omega t + 120^\circ) + 3 \cos(\omega t + 240^\circ)$.

Hard: Use cosine addition formula \rightarrow mess. If do it right, get $x(t) = 0$ (!)

Easy: Phasors: $X = 3e^{j0} + 3e^{j120^\circ} + 3e^{j240^\circ} = 0 \rightarrow x(t) = \text{Re}[Xe^{j\omega t}] = 0!$

Why? Draw picture in complex plane: easy to see resultant of these = 0!

EX#2: Show that $5 \cos(\omega t + 53^\circ) + \sqrt{2} \cos(\omega t + 45^\circ) = 6.4 \cos(\omega t + 51^\circ)$.

Soln: $5e^{j53^\circ} + \sqrt{2}e^{j45^\circ} = (3 + j4) + (1 + j) = (4 + j5) = 6.4e^{j51^\circ}$. QED.

Note: $e^{\pm j\pi} = -1$. $j = e^{j\pi/2}$ and $-j = e^{j3\pi/2} \Leftrightarrow \cos(\omega t \pm 90^\circ) = \mp \sin(\omega t)$.

Idea: Suppose voltage across and current through device are both sinusoids.
What: *Gain* and *phase shift* in going from current to voltage (and vice-versa).

Phasors: $i(t) = I_o \cos(\omega t) \Leftrightarrow I = I_o$ and $v(t) = V_o \cos(\omega t + \theta) \Leftrightarrow V = V_o e^{j\theta}$.

DEF: Impedance $Z = \frac{V}{I} = \frac{V_o}{I_o} e^{j\theta}$; **Admittance** $Y = \frac{1}{Z} = \frac{I}{V} = \frac{I_o}{V_o} e^{-j\theta}$.

Point: Can apply circuit analysis techniques to Z and Y, not just R and G.

Note: $e^{j\omega t}$ cancels throughout—don't even bother writing it at all!

Device name :	Resistor	Inductor	Capacitor
Its formula :	$v(t) = Ri(t)$	$v(t) = L \frac{di}{dt}$	$v(t) = \frac{1}{C} \int_{-\infty}^t i(\tau) d\tau$
Current $i(t)$:	$I_o \cos(\omega t)$	$I_o \cos(\omega t)$	$I_o \cos(\omega t)$
Voltage $v(t)$:	$RI_o \cos(\omega t)$	$-\omega LI_o \sin(\omega t)$	$\frac{I_o}{\omega C} \sin(\omega t)$
Gain; phase :	$R; 0$	$\omega L; +90^\circ$	$\frac{1}{\omega C}; -90^\circ$
Impedance Z :	R	$j\omega L$	$\frac{1}{j\omega C} = \frac{-j}{\omega C}$
Admittance Y :	$\frac{1}{R}$	$\frac{1}{j\omega L} = \frac{-j}{\omega L}$	$j\omega C$
Z at DC ($\omega = 0$) :	R	0	∞

DEF: $Z = R + jX$ where R=resistance and X=reactance (in Ω). $G \neq 1/R$.

DEF: $Y = G + jB$ where G=conductance and B=susceptance. $B \neq 1/X$.

Note: Impedances in series add; admittances in parallel add. $Z_1 || Z_2 = \frac{Z_1 Z_2}{Z_1 + Z_2}$.

EX: Capacitors in parallel: $Y = j\omega C_1 + \dots + j\omega C_N = j\omega(C_1 + \dots + C_N)$.

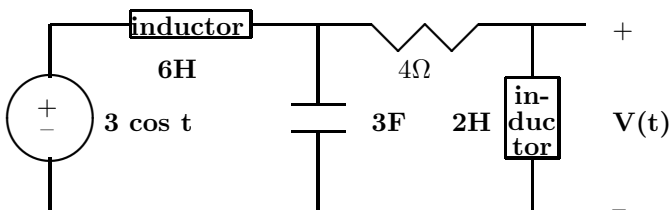
- **Phasors** are complex nos. that represent voltages, currents and sources.
- **Impedances** are nos. that represent resistors, inductors, capacitors.
- **Circuit analysis** includes KVL, KCL, node eqns, Thevenin/Norton.

EX: Illustrate various circuit techniques in the phasor domain:

1. Take **Thevenin equivalent** of everything left of the 2H inductor:

$$V_{OC} = 3 \frac{1/(j3)}{j6+1/(j3)} = \frac{3}{1-18} = -\frac{3}{17}$$

$$R_{EQ} = \frac{(j6)/(j3)}{j6+1/(j3)} + 4 = 4 - j\frac{6}{17}$$



2. Now use **voltage divider** to compute voltage across 2H inductor:

$$V_L = -\frac{3}{17} \left[\frac{j2}{j2+(4-j6/17)} \right] = \frac{-j6}{68+j28} = \frac{6e^{-j90^\circ}}{75e^{j20^\circ}} = 0.080e^{-j110^\circ}$$

3. Convert from phasor to time domain: $v(t) = 0.080 \cos(t - 110^\circ)$.