

PSHS ③:  $E_{DB} = \|x_{DB}(t)\|^2 = \int x_{DB}^2(t) dt = \int \left( \int_0^T x(u) \text{sinc}(t-u) du \right)^2 dt$  NOW WHAT?

$E_{DB} = \int \left( \int_0^T x(u) \text{sinc}(t-u) du \int_0^T x(v) \text{sinc}(t-v) dv \right) dt = \int_0^T \int_0^T x(u)x(v) \left( \int_{-\infty}^{\infty} \text{sinc}(t-u) \text{sinc}(t-v) dt \right) du dv$

$x(t) =$  EIGENFUNCTION OF  $K(t,s) = \text{sinc}(t-s)$  OVER INTERVAL  $[0, T]$  ASSOCIATED WITH LARGEST  $\lambda$

OBSERVE  $r(t) = \begin{cases} s_1(t) + n(t) & H_1 \\ s_2(t) + n(t) & H_2 \end{cases}$  OVER  $0 \leq t \leq T$   $\left[ \begin{matrix} n(t) \text{ WGN} \\ s_n(t) = 0 \end{matrix} \right]$   $\left[ \begin{matrix} s_1(t) \perp s_2(t) \\ \int_0^T s_1(t)s_2(t) dt = 0 \end{matrix} \right]$   $\left[ \begin{matrix} \int_0^T s_1(t)^2 dt = E_1 \\ \int_0^T s_2(t)^2 dt = E_2 \end{matrix} \right]$

K-L:  $n(t)$  WHITE  $\rightarrow$  USE ANY C.O.S.  $\left\{ \begin{matrix} \phi_1(t) = s_1(t)/\sqrt{E_1} \\ \phi_2(t) = s_2(t)/\sqrt{E_2} \end{matrix} \right.$   $\left\{ \begin{matrix} \phi_3(t) \perp \{\phi_1, \phi_2\} \\ \phi_4(t) \perp \{\phi_1, \phi_2, \phi_3\} \end{matrix} \right.$  [GRAM-SCHMIDT]  $E(n_i^2) = \lambda_i = \sigma^2$   $E(n_i, n_j) = \sigma^2 \delta_{ij}$

PROJECT: MUT. BY  $\phi_i(t)$  THEN  $\int_0^T dt$ :  $v_1 = \begin{cases} \sqrt{E_1} + n_1 & H_1 \\ n_1 & H_2 \end{cases}$   $v_2 = \begin{cases} n_2 & H_1 \\ \sqrt{E_2} + n_2 & H_2 \end{cases}$   $v_3 = \begin{cases} n_3 & H_1 \\ n_3 & H_2 \end{cases}$   $v_4 = \begin{cases} n_4 & H_1 \\ n_4 & H_2 \end{cases}$   $n_i$ : GAUSS  $n_i$ : GRV  $n_i$ : INDEP

MULTICHANNEL GAUSSIAN:  $\begin{bmatrix} v_1 \\ v_2 \end{bmatrix} \sim \begin{cases} N \left( \begin{bmatrix} \sqrt{E_1} \\ 0 \end{bmatrix}, \sigma^2 I \right) & H_1 \\ N \left( \begin{bmatrix} 0 \\ \sqrt{E_2} \end{bmatrix}, \sigma^2 I \right) & H_2 \end{cases}$  SEE HANDOUT ON "VECTOR GAUSSIAN DETECTION" / MATCHED FILTER

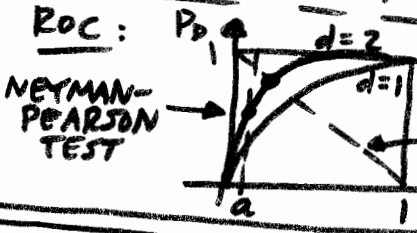
SUFFICIENT STATISTIC:  $l = \Delta m^T Q R = \left( \begin{bmatrix} \sqrt{E_1} \\ 0 \end{bmatrix} - \begin{bmatrix} 0 \\ \sqrt{E_2} \end{bmatrix} \right)^T \frac{1}{\sigma^2} I \begin{bmatrix} \int_0^T R(t) s_1(t) / \sqrt{E_1} dt \\ \int_0^T R(t) s_2(t) / \sqrt{E_2} dt \end{bmatrix} = \int_0^T R(t) (s_1(t) - s_2(t)) dt$

FISHER DISCRIMINANT  $d^2 = \Delta m^T Q \Delta m = \left( \begin{bmatrix} \sqrt{E_1} \\ 0 \end{bmatrix} - \begin{bmatrix} 0 \\ \sqrt{E_2} \end{bmatrix} \right)^T \frac{1}{\sigma^2} I \left( \begin{bmatrix} \sqrt{E_1} \\ 0 \end{bmatrix} - \begin{bmatrix} 0 \\ \sqrt{E_2} \end{bmatrix} \right) = \frac{E_1 + E_2}{\sigma^2}$  DIMENSIONLESS  $\left( \frac{d^2}{\sigma^2} \right)$  SIGNAL ENERGIES AS SINCE ORTHOGONAL.

NEYMAN-PEARSON TEST:  $P_{L|H_2}(L|H_2) \sim N(\Delta m^T Q m_2, \Delta m^T Q \Delta m) \sim N(-E_2/\sigma^2, E_1 + E_2/\sigma^2)$  NOTE  $\sigma_l^2 = d^2$ ; SEE HANDOUT

$(H_2 = \text{NULL})$   $\alpha = P_F = P_r[\text{CHOOSE } H_1 | H_2 \text{ TRUE}] = P_r[l > \gamma | H_2] = \text{erfc} \left[ \frac{\gamma + E_2/\sigma^2}{\sqrt{(E_1 + E_2)/\sigma^2}} \right] = \alpha$

SOLVE FOR  $\gamma$ :  $\gamma = -\frac{E_2}{\sigma^2} + \sqrt{\frac{E_1 + E_2}{\sigma^2}} \text{erfc}^{-1}[\alpha]$



SAY  $d^2 = \frac{E_1 + E_2}{\sigma^2} = 4$   
 $\rightarrow d = 2$   
 MINIMAX TEST WITH MEP CRITERION

BAYESIAN:  $\int_0^T R(t) (s_1(t) - s_2(t)) dt \underset{H_1}{\geq} \frac{E_1 - E_2 + \sigma \sqrt{E_1 + E_2}}{2} \text{erfc}^{-1}[\alpha]$   
 $\int_0^T R(t) (s_1(t) - s_2(t)) dt \underset{H_0}{\geq} \frac{E_1 - E_2 + \sigma^2 \log \eta}{2}$

CAN ALSO PERIVE THIS WITHOUT USING RESULTS FROM "VECTOR GAUSSIAN DETECTION":

UNDER  $H_1$ :  $l = \int_0^T R(t) (s_1(t) - s_2(t)) dt = \int_0^T (s_1(t) + n(t)) (s_1(t) - s_2(t)) dt = \int_0^T s_1^2(t) dt - \int_0^T s_1(t)s_2(t) dt + \int_0^T (s_1(t) - s_2(t)) n(t) dt = E_1 + \int_0^T (s_1(t) - s_2(t)) n(t) dt$

$\sigma_l^2 = \sigma^2 \left[ E_1 + \int_0^T (s_1(t) - s_2(t)) n(t) dt \right]^2 = E \left[ \left( \int_0^T (s_1(t) - s_2(t)) n(t) dt \right)^2 \right]$  NOW WHAT?  
 $= E \left[ \int_0^T \int_0^T (s_1(u) - s_2(u)) n(u) (s_1(v) - s_2(v)) n(v) du dv \right]$   
 $= \int_0^T \int_0^T (s_1(u) - s_2(u)) (s_1(v) - s_2(v)) E(n(u)n(v)) du dv = \sigma^2 \int_0^T (s_1(u) - s_2(u))^2 du = \sigma^2 (E_1 + E_2)$  SINCE  $s_1 \perp s_2$

$\therefore P_{L|H_1}(L|H_1) \sim N(E_1, \sigma^2(E_1 + E_2))$  SIMILARLY, UNDER  $H_2$ :  $l = \int_0^T (s_2(t) + n(t)) (s_1(t) - s_2(t)) dt = \int_0^T s_2 s_1 - \int_0^T s_2^2 + \int_0^T (s_1 - s_2) n(t) dt$  AS BEFORE.

NOTE:  $l$  DEFINED DIFFERENTLY: NO  $\sigma^2$ .

GENERAL BINARY DETECTION IN WHITE NOISE

**PROBLEM** : OBSERVE  $r(t) = \begin{cases} s_1(t) + w(t) & \text{UNDER } H_1 \\ s_2(t) + w(t) & \text{UNDER } H_2 \end{cases}$   $\int_0^T s_1^2(t) dt = E_1$ , FOR  $0 \leq t \leq T$ .  
 $\int_0^T s_2^2(t) dt = E_2$   $w(t)$  WGN, STRENGTH  $\sigma^2$ .

**CAUCHY-SCHWARZ**  $\neq$  :  $(\int_0^T s_1(t) s_2(t) dt)^2 \leq (\int_0^T s_1^2(t) dt) (\int_0^T s_2^2(t) dt) = E_1 E_2$ .  
 DEFINE  $\cos \theta = \frac{\int_0^T s_1(t) s_2(t) dt}{\sqrt{E_1} \sqrt{E_2}}$  (KNOWN) WHERE  $0 \leq \theta \leq 180^\circ$

**BASIS FUNCTIONS** :  $\phi_1(t) = \frac{s_1(t)}{\sqrt{E_1}}$ ,  $\phi_2(t) = \frac{s_2(t) - \phi_1(t) (\int_0^T s_2(t) \phi_1(t) dt)}{\sqrt{\int_0^T [s_2(t) - \phi_1(t) (\int_0^T s_2(t) \phi_1(t) dt)]^2 dt}}$  (GRAM-SCHMIDT).

Now,  $\int_0^T s_2(t) \phi_1(t) dt = \int_0^T \frac{s_1(t) s_2(t)}{\sqrt{E_1}} dt = \sqrt{E_2} \cos \theta$  (\*)

**DENOMINATOR OF  $\phi_2(t)$**  =  $\sqrt{\int_0^T s_2^2(t) dt + \int_0^T \phi_1^2(t) (E_2 \cos^2 \theta) - 2 \int_0^T s_2(t) \phi_1(t) \sqrt{E_2} \cos \theta} = \sqrt{E_2 (1 - \cos^2 \theta)} = \sqrt{E_2} \sin \theta$ .  
 SINCE  $0 \leq \theta \leq 180^\circ$

$\therefore \phi_2(t) = \frac{s_2(t) - \phi_1(t) \sqrt{E_2} \cos \theta}{\sqrt{E_2} \sin \theta} = \frac{s_2(t) - \sqrt{\frac{E_2}{E_1}} \cos \theta s_1(t)}{\sqrt{E_2} \sin \theta}$ . IF  $\theta = 90^\circ$ ,  $\phi_2(t) = \frac{s_2(t)}{\sqrt{E_2}}$  ✓

NOW LET  $r(t) = \sum_{i=1}^{\infty} v_i \phi_i(t)$ ,  $w(t) = \sum_{i=1}^{\infty} w_i \phi_i(t)$ .  $\phi_1(t), \phi_2(t)$  AS ABOVE.  $\{\phi_3, \dots\} \perp \phi_1(t), \phi_2(t)$ .  
 NOTE THAT SINCE  $r(t) = w(t) +$  DETERMINISTIC SIGNAL,  $r(t)$  AND  $w(t)$  HAVE SAME BASIS FUNCTIONS.  
 NOW TAKE PROJECTIONS ON  $\phi_i(t)$  BY MULT.  $r(t) = \begin{cases} s_1(t) + w(t) & H_1 \\ s_2(t) + w(t) & H_2 \end{cases}$  BY  $\phi_i(t)$  AND  $\int_0^T dt$ .

LOOK AT  $i$ TH COORDINATES

$i=1$  :  $v_1 = \begin{cases} \int_0^T s_1(t) \phi_1(t) dt + w_1 \\ \int_0^T s_2(t) \phi_1(t) dt + w_1 \end{cases} = \begin{cases} \sqrt{E_1} + w_1 & H_1 \\ \sqrt{E_2} \cos \theta + w_1 & H_2 \end{cases}$   $w(t) = \sum_{i=1}^{\infty} w_i \phi_i(t)$ ,  $E(w_i^2) = \sigma^2$ .

$i=2$  :  $v_2 = \begin{cases} \int_0^T s_1(t) \phi_2(t) dt + w_2 \\ \int_0^T s_2(t) \phi_2(t) dt + w_2 \end{cases} = \begin{cases} 0 (\phi_2(t) \perp s_1(t) = \sqrt{E_1} \phi_1(t)) + w_2 \\ \int_0^T s_2^2(t) dt - \int_0^T s_2(t) \phi_1(t) \sqrt{E_2} \cos \theta + w_2 \end{cases} = \begin{cases} w_2 & H_1 \\ \sqrt{E_2} \sin \theta + w_2 & H_2 \end{cases}$

$i > 2$  :  $v_i = \begin{cases} w_i & H_1 \\ w_i & H_2 \end{cases}$  USELESS.

$\therefore [v_1, v_2]^T \sim N \left( \begin{bmatrix} \sqrt{E_1} \\ 0 \end{bmatrix}, \begin{bmatrix} \sigma^2 & 0 \\ 0 & \sigma^2 \end{bmatrix} \right)$  UNDER  $H_1$ ;  $[v_1, v_2]^T \sim N \left( \begin{bmatrix} \sqrt{E_2} \cos \theta \\ \sqrt{E_2} \sin \theta \end{bmatrix}, \begin{bmatrix} \sigma^2 & 0 \\ 0 & \sigma^2 \end{bmatrix} \right)$  UNDER  $H_2$ .

**TEST IS**  $l = \underline{a}^T \underline{v} = (\sqrt{E_1} - \sqrt{E_2} \cos \theta) R_1 + (0 - \sqrt{E_2} \sin \theta) R_2$   $\frac{H_1}{\geq} \gamma$ . BUT  $R_1 = \int_0^T r(t) \phi_1(t) dt$ ,  $R_2 = \int_0^T r(t) \phi_2(t) dt$ .

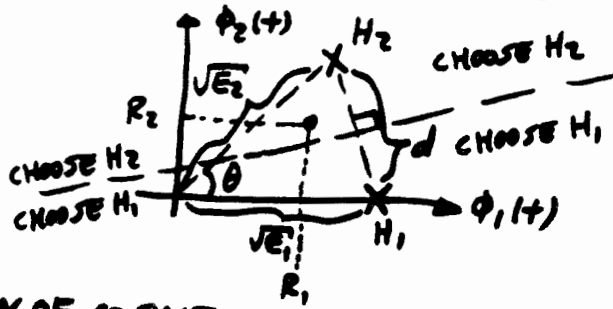
SUBSTITUTING,  $l = \int_0^T r(t) \sqrt{E_1} \phi_1(t) dt - \sqrt{E_2} \cos \theta \int_0^T r(t) s_1(t) dt - \int_0^T r(t) (\sqrt{E_2} \sin \theta \phi_2(t)) dt$   
 $\rightarrow l = \int_0^T r(t) (s_1(t) - s_2(t)) dt \stackrel{H_1}{\geq} \gamma$ .  
 $= \int_0^T r(t) [s_2(t) - \sqrt{\frac{E_2}{E_1}} \cos \theta s_1(t)] dt$

THAT IS,  $r(t) = \begin{cases} s_1(t) + w(t) & H_1 \\ s_2(t) + w(t) & H_2 \end{cases} \rightarrow r'(t) = r(t) - s_2(t) = \begin{cases} (s_1(t) - s_2(t)) + w(t) & H_1 \\ w(t) & H_2 \end{cases}$   
 $\therefore$  MATCHED FILTER TO  $s_1(t) - s_2(t)$ . (KNOWN DETERMINISTIC)

PERFORMANCE OF TEST IS  $d^2 = \Delta \underline{m}^T \Theta \Delta \underline{m} = \frac{1}{\sigma^2} [(\sqrt{E_1} - \sqrt{E_2} \cos \theta)^2 + (0 - \sqrt{E_2} \sin \theta)^2]$

$$d^2 = \frac{1}{\sigma^2} (E_1 + E_2 \cos^2 \theta - 2\sqrt{E_1}\sqrt{E_2} \cos \theta + E_2 \sin^2 \theta) = \frac{1}{\sigma^2} (E_1 + E_2 - 2\sqrt{E_1}\sqrt{E_2} \cos \theta)$$

SIGNAL SPACE INTERPRETATION:



BY LAW OF COSINES,

$$d^2 = E_1 + E_2 - 2\sqrt{E_1}\sqrt{E_2} \cos \theta$$

THEN SCALE BY  $\sigma^2$ .

$R(t) = R_1 \phi_1(t) + R_2 \phi_2(t) + \dots$   $R_i = \int_0^T R(t) \phi_i(t) dt$   
 COORDINATES:  $[R_1, R_2, R_3, \dots]$   
 $S_1(t) = \sqrt{E_1} \phi_1(t) + 0 \cdot \phi_2(t)$   
 COORDINATES:  $[\sqrt{E_1}, 0, 0, \dots]$   
 $S_2(t) = (\sqrt{E_2} \cos \theta) \phi_1(t) + (\sqrt{E_2} \sin \theta) \phi_2(t)$   
 COORDINATES:  $[\sqrt{E_2} \cos \theta, \sqrt{E_2} \sin \theta, 0, \dots]$   
 COMPUTE THESE USING  $i=1, i=2$   
 ON OTHER SIDE, BUT USING SIGNAL SPACES,  
 CAN SEE WHY THIS IS TRUE!

NOTE THAT ONLY THE 1<sup>ST</sup> 2 COORDINATES  $R_1$  AND  $R_2$  OF  $R(t)$  MATTER:

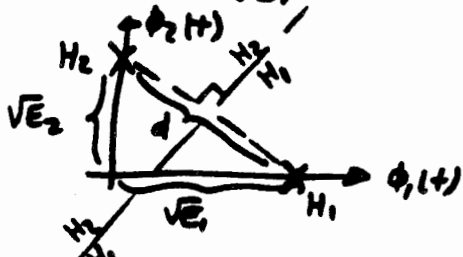
PROJECT  $R(t)$  ONTO  $\text{SPAN}(\phi_1(t), \phi_2(t))$   
 $\rightarrow (R_1, R_2)$  IN  $\phi_1(t)$ - $\phi_2(t)$  PLANE.

IF  $\theta = 90^\circ \rightarrow \int_0^T S_1(t) S_2(t) dt = 0$  EX:  $S_1(t) = \sin 2\pi t$ ,  $S_2(t) = \sin 2\pi t + t$

NOW  $\phi_1(t) = \frac{S_1(t)}{\sqrt{E_1}}$ ,  $\phi_2(t) = \frac{S_2(t)}{\sqrt{E_2}}$

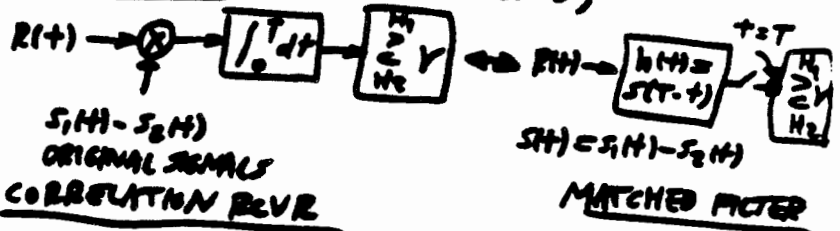
$$d^2 = \frac{E_1 + E_2}{\sigma^2}$$

SET  $\theta = 90^\circ$  IN ABOVE EQNS.



BY PYTHAGORAS,  $d^2 = E_1 + E_2$   
 THEN SCALE BY  $\sigma^2$ .

SUMMARY: REGARDLESS OF  $\theta$ ,



PERFORMANCE:  $d^2 = (E_1 + E_2 - 2\sqrt{E_1}\sqrt{E_2} \cos \theta) / \sigma^2$   
 USE TO PFK ROC CURVE ON P. 38  
 (IN GENERAL:  $P_D = \text{erfc}[s]$ ,  $P_D = \text{erfc}[s+d]$   
 AS  $s$  VARIES FROM  $-\infty$  TO  $\infty$ )

**PROBLEM:** OBSERVE  $r(t) = \begin{cases} \sqrt{E_i} s_i(t) + w(t) & \text{UNDER } H_i, \text{ FOR } 0 \leq t \leq T \\ \sqrt{E_M} s_M(t) + w(t) & \text{UNDER } H_M \end{cases}$  FOR  $0 \leq t \leq T$   
 $w(t) = 0$ -MEAN WGN WITH STRENGTH  $\frac{N_0}{2}$

**GOAL:** GIVEN  $\{R_i(t), 0 \leq t \leq T\}$ ,  $\int_0^T s_i(t) s_j(t) dt = \rho_{ij}, 1 \leq i, j \leq M$ .  
 CHOOSE ONE OF  $H_1, H_2, \dots, H_M$ .

**CRITERION:** MIN  $P_V$  [ERROR].  
 NOTE  $|\rho_{ij}| \leq \rho_{ii} = 1$  WITHOUT LOSS OF GENERALITY.

**THE FOLLOWING ARE GIVEN:**  $s_i(t), E_i, \rho_{ij}, N_0/2, \rho_i = P_V[H_i \text{ TRUE}]$  (A PRIORI PROBS)

**SOLUTION:** (1) PERFORM GRAM-SCHMIDT ON  $\{s_1(t) \dots s_M(t)\} \rightarrow \{\phi_1(t) \dots \phi_N(t)\}$

USING 
$$\phi_n(t) = \frac{s_n(t) - \sum_{i=1}^{n-1} \phi_i(t) \left( \int_0^T s_n(t) \phi_i(t) dt \right)}{\sqrt{\int_0^T \left( s_n(t) - \sum_{i=1}^{n-1} \phi_i(t) \left( \int_0^T s_n(t) \phi_i(t) dt \right) \right)^2 dt}}$$
 FOR  $n=1, 2, \dots, N$

**NOTE:** ALMOST SURELY  $N=M$ . HOWEVER, IF  $\{s_1(t) \dots s_M(t)\}$  IS NOT LINEARLY INDEPT, THEN  $N < M$ .

(2) USE K-L EXPANSION:

E.G.,  $s_M(t) = \sum_{i=1}^M s_i(t) \rightarrow$  CAN ONLY GET  $N=M-1$  BASIS FUNCS.

DEFINE  $m_{ij} = \sqrt{E_j} \int_0^T s_j(t) \phi_i(t) dt, 1 \leq i \leq N, 1 \leq j \leq M$

MULTIPLY "OBSERVE" EQN BY  $\phi_i(t)$  AND  $\int_0^T dt$ :

$$r_i = \int_0^T r(t) \phi_i(t) dt = \begin{cases} m_{i1} + w_i & H_1 \\ m_{i2} + w_i & H_2 \\ \vdots & \vdots \\ m_{iM} + w_i & H_M \end{cases} = \begin{cases} w_i & H_1 \\ w_i & H_2 \\ \vdots & \vdots \\ w_i & H_M \end{cases} \quad \begin{matrix} 1 \leq i \leq N \\ i > N \text{ (NO HELP)} \end{matrix}$$

(3) APPLY M-ARY HYPOTHESIS TESTING RESULTS FROM CHAP. 3:

CHOOSE  $H_i$  WHERE  $\sum_{j \neq i} (c_{ij} - c_{jj}) \rho_j P_{r_i|H_j} (R_i|H_j)$  SMALLEST  $\rightarrow$  CHOOSE  $H_i$  WHERE  $P_{r_i|H_i} (R_i|H_i) \rho_i$  LARGEST

[GENERAL RESULT] FROM ABOVE,  $P_{r_i|H_i} (R_i|H_i) \sim N(m_{ii}, N_0/2)$  [MPP:  $c_{ij} = 1 - \delta_{ij}$ ] SINCE  $w_j \sim N(0, N_0/2)$  AND  $P_{r_i|H_i} (R_i|H_i) = \prod_{j=1}^N P_{r_j|H_i} (R_j|H_i)$  [MIN  $P_V$  (ERROR) RESULT]

$\rightarrow$  CHOOSE  $H_i$  WHERE  $\left( \frac{1}{(2\pi N_0/2)^N} \prod_{j=1}^N e^{-\frac{(R_j - m_{ji})^2}{2 \cdot N_0/2}} \right) \rho_i$  LARGEST [PLUG IN] SINCE  $\{w_j\}$  UNCORRELATED GAUSSIANS  $\rightarrow$  INDEP RVs.

$\rightarrow$  CHOOSE  $H_i$  WHERE  $\log \rho_i - \frac{1}{N_0} \sum_{j=1}^N (R_j - m_{ji})^2$  LARGEST AND  $R_j = \int_0^T r(t) \phi_j(t) dt$  [TAKE LOG]

**FINAL PROCEDURE:**

- (1)  $\{s_1(t) \dots s_M(t)\} \xrightarrow{\text{GRAM SCHMIDT}} \{\phi_1(t) \dots \phi_N(t)\}$  (2) COMPUTE  $m_{ij} = \sqrt{E_j} \int_0^T s_j(t) \phi_i(t) dt, 1 \leq i \leq N, 1 \leq j \leq M$  [NOTE  $m_{ij} = 0$  IF  $i > j$ ]
- (3) READ IN DATA  $\{r(t), 0 \leq t \leq T\}$ . COMPUTE  $R_j = \int_0^T r(t) \phi_j(t) dt, 1 \leq j \leq N$ .
- (4) COMPUTE  $\log \rho_i - \frac{1}{N_0} \sum_{j=1}^N (R_j - m_{ji})^2$  FOR  $1 \leq i \leq M$ . (5) CHOOSE  $H_k$  WHERE  $i=k$  MAXIMIZES (4).

**SPECIAL CASE #1:** EACH HYPOTHESIS EQUALLY LIKELY A PRIORI:  $\rho_i = \frac{1}{M}, 1 \leq i \leq M$ . EQUAL ENERGIES:  $E_j = E$ . ALSO, SIGNALS UNCORRELATED:  $\int_0^T s_i(t) s_j(t) dt = \rho_{ij} = \delta_{ij}$ . THEN  $\phi_i(t) = s_i(t)$

$\rightarrow \log \rho_i - \frac{1}{N_0} \sum_{j=1}^N (R_j - m_{ji})^2 = \log \frac{1}{M} - \frac{1}{N_0} \left( \sum_{j=1}^N R_j^2 + E - 2 \sum_{j=1}^N R_j \sqrt{E} \delta_{ij} \right)$  [INDEP OF I] [INDEP OF I] ALSO: SEE OVERLEAF  $m_{ij} = \sqrt{E} \delta_{ij}$

$\rightarrow$  CHOOSE  $H_i$  WHERE  $R_i$  LARGEST AND  $R_i = \int_0^T r(t) s_i(t) dt = R_i$  HERE. [NOTE  $\frac{1}{N_0}$  = TRADEOFF FACTOR BETWEEN A PRIORI  $\log \rho_i$  AND A POSTERIORI  $\sum (R_j - m_{ji})^2$ ]

REMARKABLY, THE ABOVE RESULT HOLDS EVEN IF THE SIGNALS ARE CORRELATED:

SPECIAL CASE #2: IF EACH HYPOTHESIS EQUALLY LIKELY A PRIORI ( $P_i = \frac{1}{M}$ )  
AND IF EACH SIGNAL HAS THE SAME ENERGY  $E$  ( $E_i = E$  FOR  $1 \leq i \leq M$ )

THEN THE OPTIMAL RECEIVER IS: **CHOOSE  $H_i$  WHERE  $R_i = \int_0^T R_i(t)S_i(t)dt$  IS LARGEST**  
EVEN IF THE SIGNALS ARE CORRELATED!!

PROOF: RECALL PARSEVAL'S THM FOR REAL ORTHONORMAL EXPANSIONS:

IF  $R(t) = \sum_{i=1}^N R_i \phi_i(t)$  AND  $S(t) = \sum_{i=1}^N S_i \phi_i(t)$ , THEN  $\int_0^T R(t)S(t)dt = \sum_{i=1}^N R_i S_i$ .

THEN  $\log P_i = \frac{1}{N_0} \sum_{j=1}^N (R_j - m_{j,i})^2 = \log \frac{1}{M} - \frac{1}{N_0} \sum_{j=1}^N R_j^2 - \frac{1}{N_0} \sum_{j=1}^N m_{j,i}^2 + \frac{2}{N_0} \sum_{j=1}^N R_j m_{j,i}$ .

BUT  $m_{j,i} = \int_0^T S_i(t) \phi_j(t) dt \rightarrow S_i(t) = \sum_{j=1}^N m_{j,i} \phi_j(t)$  AND  $R_j = \int_0^T R(t) \phi_j(t) dt \rightarrow R(t) = \sum_{j=1}^N R_j \phi_j(t)$ .

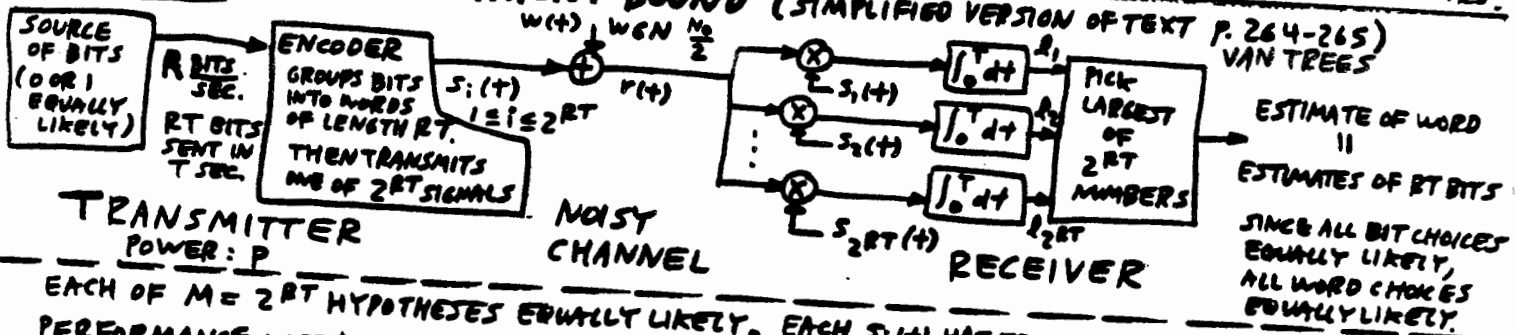
SO  $-\frac{1}{N_0} \sum_{j=1}^N m_{j,i}^2 = -\frac{1}{N_0} \int_0^T S_i(t)^2 dt = -\frac{1}{N_0} E$ , LEAVING  $\frac{2}{N_0} \sum_{j=1}^N R_j m_{j,i} = \frac{2}{N_0} \int_0^T R(t)S_i(t)dt$ .

$\therefore$  CHOOSE  $H_i$  WHERE  $\log P_i - \frac{1}{N_0} \sum_{j=1}^N (R_j - m_{j,i})^2$  LARGEST  $\rightarrow$  CHOOSE  $H_i$  WHERE  $\int_0^T R(t)S_i(t)dt$  LARGEST.

NOTE: THIS ISN'T ALL THAT SURPRISING -- WE SAW IT BEFORE IN THE CASE OF GENERAL BINARY DETECTION. EVEN WHEN  $S_1(t)$  AND  $S_2(t)$  WERE CORRELATED, THE OPTIMAL TEST WAS  $\int_0^T R(t)(S_1(t) - S_2(t))dt \stackrel{H_1}{\geq} \frac{Y}{2} = 0$  BY SYMMETRY IF SAME ENERGIES AND EQUAL A PRIORI PROBS.  $\rightarrow \int_0^T R(t)S_1(t)dt \stackrel{H_1}{\geq} \int_0^T R(t)S_2(t)dt$  WHICH AGREES WITH ABOVE. SEE TEXT P. 257

WHILE FORM OF OPTIMAL TEST IS UNAFFECTED ABOVE BY CORRELATION BETWEEN SIGNALS, PERFORMANCE OF OPTIMAL TEST IS AFFECTED. BEST PERFORMANCE: USE ORTHOGONAL SIGNALS.

EXAMPLE: CHANNEL CAPACITY BOUND (SIMPLIFIED VERSION OF TEXT P. 264-265) VAN TREES



EACH OF  $M = 2^{RT}$  HYPOTHESES EQUALLY LIKELY. EACH  $S_i(t)$  HAS EQUAL ENERGY  $PT$ .  
PERFORMANCE: LET'S COMPUTE A BOUND ON  $P_r$  (ERROR). BY SYMMETRY, WLOG WE CAN LET  $H_1$  BE TRUE.

$$P_r(\text{ERROR}) = P_r[l_2 > l_1, \text{ OR } l_3 > l_1, \text{ OR } \dots \text{ OR } l_{2^{RT}} > l_1] \leq (2^{RT} - 1) P_r[l_i > l_1] = (2^{RT} - 1) \text{erfc} \left[ \sqrt{\frac{PT}{N_0}} \right]$$

[WAYS CHOICE OF  $l_i$  CAN BE MADE] [UNION BOUND] [EQUAL BY SYMMETRY]

FROM (7D) ON P. 39, WE HAVE  $\text{erfc}(x) < \frac{1}{2} e^{-x^2/2}$  (A VERY CRUDE BOUND) [VAN TREES] [170] ON P. 38

$$\rightarrow P_r(\text{ERROR}) \leq 2^{RT} \text{erfc} \left[ \sqrt{\frac{PT}{N_0}} \right] < 2^{RT} \frac{1}{2} e^{-PT/2N_0} = \frac{1}{2} e^{(RT \log 2 - PT/2N_0)} = \frac{1}{2} 2^{T(R - P/N_0 \log 2)}$$

[UNION BOUND] [erfc BOUND] [2 = e^{log 2}]

NOW LET  $T \rightarrow \infty$  (NOTE THAT  $M = 2^{RT} = \text{BANDWIDTH} \rightarrow \infty$  ALSO). WHAT HAPPENS TO THIS BOUND?

IF  $R < P/2N_0 \log 2$ , THEN  $T(R - P/2N_0 \log 2) < 0$  AND  $\frac{1}{2} 2^{T(R - P/2N_0 \log 2)} \rightarrow 0$  SO  $P_r(\text{ERROR}) \rightarrow 0$ .

IF  $R > P/2N_0 \log 2$ , THEN  $\frac{1}{2} 2^{T(R - P/2N_0 \log 2)} \rightarrow \infty$ ! ALTHOUGH THIS IS ONLY AN UPPER BOUND ON  $P_r(\text{ERROR})$ .  
ACTUAL CHANNEL CAPACITY (SHANNON):  $R < P/N_0 \log 2$ . WHY OFF BY 2? TOO MANY BOUNDS. THRESHOLD! THIS IS NOT A GOOD SIGN!